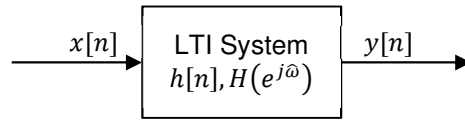


PROBLEM 1: Parts a and b can be solved independently of each other

(a) Consider the LTI system shown below with input $x[n]$ and output $y[n]$, whose frequency response is $H(e^{j\hat{\omega}}) = 10e^{-j26\hat{\omega}}(10 + 10\cos(10\hat{\omega}))$:



(i.) Find the **difference equation** that relates $y[n]$ to $x[n]$. (5 points)

$$H(e^{j\hat{\omega}}) = 10e^{-j26\hat{\omega}}(10 + 10\cos(10\hat{\omega})) = 100e^{-j26\hat{\omega}} + 50e^{-j16\hat{\omega}} + 50e^{-j36\hat{\omega}}$$

$$\rightarrow h[n] = 50\delta[n - 16] + 100\delta[n - 26] + 50\delta[n - 36]$$

$$y[n] = 50x[n - 16] + 100x[n - 26] + 50x[n - 36]$$

(ii.) If $x[n] = 2 + 2\cos\left(\frac{\pi}{5}n\right)$ find an equation for the output $y[n]$. (Hint: It has the form $y[n] = A + B\cos(\hat{\omega}_0n + \phi)$). (5 points)

$$H(e^{j0}) = 200; H\left(e^{j\frac{\pi}{5}}\right) = 200e^{-j\frac{26\pi}{5}} \rightarrow y[n] = 400 + 400\cos\left(\frac{\pi}{5}n - \frac{26\pi}{5}\right) = 400 + 400\cos\left(\frac{\pi}{5}n + \frac{4\pi}{5}\right)$$

$$y[n] = 400 + 400\cos\left(\frac{\pi}{5}n + \frac{4\pi}{5}\right)$$

(b) Consider the impulse response $h[n] = \frac{1}{5}(u[n - 20] - u[n - 33])$. **Use the provided tables of DTFT pairs and properties** to find the frequency response $H(e^{j\hat{\omega}})$ and determine the magnitude of the frequency response at $\hat{\omega} = 0$ (i.e., $H(e^{j0})$). (NOTE: You must express the frequency response in the form $H(e^{j\hat{\omega}}) = A(\hat{\omega})e^{j\varphi(\hat{\omega})}$, where $A(\hat{\omega})$ presents the overall amplitude function and $\varphi(\hat{\omega})$ represents the overall phase function of the frequency response) (10 points)

This is a “box” in discrete time of length 13 samples shifted by 20 samples.

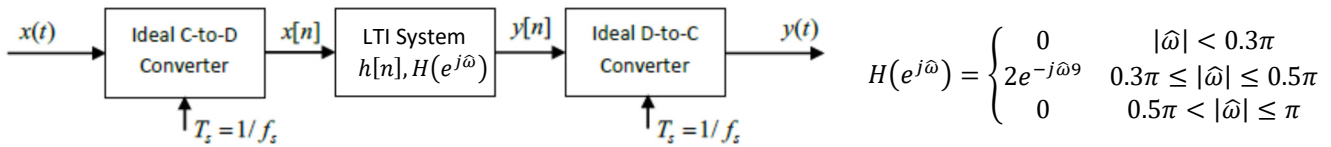
$$H(e^{j\hat{\omega}}) = \frac{\sin\left(\frac{13\hat{\omega}}{2}\right)}{5\sin\left(\frac{\hat{\omega}}{2}\right)}e^{-j\hat{\omega}26} \rightarrow H(e^{j0}) = \frac{13}{5}$$

$$H(e^{j\hat{\omega}}) = \frac{\sin\left(\frac{13\hat{\omega}}{2}\right)}{5\sin\left(\frac{\hat{\omega}}{2}\right)}e^{-j\hat{\omega}26}$$

$$H(e^{j0}) = \frac{13}{5}$$

PROBLEM 2: All parts can be solved independently of each other

Use the following ideal C-to-D / D-to-C system with the defined LTI System $H(e^{j\hat{\omega}})$ for all parts below.



Let $x(t) = 9 + 3 \cos(900\pi t) + 6 \cos(1100\pi t)$

(a) If $f_s = 2000 \text{ Hz}$, find the output $y(t)$. (12 points)

$$x[n] = 9 + 2 \cos(0.45n) + 6 \cos(0.55\pi n)$$

$$y[n] = 0 * 9 + 2 * 3 \cos(0.45\pi n - 4.05\pi) + 0 * 6 \cos(0.55\pi n - 4.95\pi) \rightarrow 4 \cos(0.45\pi n - 0.05\pi)$$

$$y(t) = 6 \cos(900\pi t - 0.05\pi)$$

$$y(t) = 6 \cos(900\pi t - 0.05\pi)$$

(b) For what range of sampling rates is there both **no aliasing** and a **nonzero output** $y(t)$. Your answer should be expressed as $A \leq f_s \leq B$ where you must determine A and B . (Hint: a nonzero output for $y(t)$ means that a minimum of one of the components of $x(t) = 9 + 3 \cos(900\pi t) + 6 \cos(1100\pi t)$ is passed by the LTI system) (8 points)

For $x(t)$ has frequencies of 900π and 1100π . This problem requires (1) No aliasing (2) At least one digital frequency in the range $0.3\pi \leq |\hat{\omega}| \leq 0.5\pi$.

(1) No aliasing: $f_s > 1100 \text{ Hz}$

(2) At least one component in the range $0.3\pi \leq |\hat{\omega}| \leq 0.5\pi$

$$\begin{aligned} \rightarrow \frac{900\pi}{f_s} \leq 0.5\pi \rightarrow f_s \geq 1800; \frac{900\pi}{f_s} \geq 0.3\pi \rightarrow f_s \leq 3000 \\ \rightarrow \frac{1100\pi}{f_s} \leq 0.5\pi \rightarrow f_s \geq 2200; \frac{1100\pi}{f_s} \geq 0.3\pi \rightarrow f_s \leq \frac{1100}{0.3} \approx 3666.67 \end{aligned}$$

Combining the ranges yields: $1800 \leq f_s \leq \frac{1100}{0.3} \approx 3666.67$

$$1800 \leq f_s \leq \frac{1100}{0.3} \approx 3666.67$$

PROBLEM 3: All parts can be solved independently of each other

(a) An LTI system with output $y[n]$ has impulse response $h[n] = 4\delta[n - 1] + 3\delta[n - 3] - 4\delta[n - 7]$. If the input to the system is the unit step function $x[n] = u[n]$, find $y[4]$ and $y[9]$.

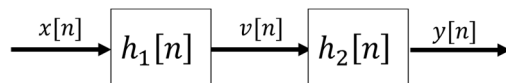
(10 points) (NOTE: You don't need to make a table)

$$y[n] = 4u[n - 1] + 3u[n - 3] - 4u[n - 7] \rightarrow y[4] = 4u[3] + 3u[1] - 4u[-3] = 7$$

$$y[9] = 4u[8] + 3u[6] - 4u[2] = 3$$

$y[4] = 7$	$y[9] = 3$
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(b) Consider the cascaded LTI system below



where $h_1[n] = \sum_{k=3}^8 \delta[n - k]$ and $h_2[n] = \delta[n] - \delta[n - 2]$. Find an equation for $y[n]$ that is valid for all n , when $x[n] = 2\delta[n - 3]$. (10 points)

Overall impulse response is: $h[n] = \delta[n - 3] + \delta[n - 4] - \delta[n - 9] - \delta[n - 10]$

$$y[n] = 2\delta[n - 6] + 2\delta[n - 7] - 2\delta[n - 12] - 2\delta[n - 13]$$

$y[n] = 2\delta[n - 6] + 2\delta[n - 7] - 2\delta[n - 12] - 2\delta[n - 13]$
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Table of DTFT Pairs	
<i>Time-Domain: $x[n]$</i>	<i>Frequency-Domain: $X(e^{j\hat{\omega}})$</i>
$\delta[n]$	1
$\delta[n - n_d]$	$e^{-j\hat{\omega}n_d}$
$r_L[n] = u[n] - u[n - L]$	$\frac{\sin(\frac{1}{2}L\hat{\omega})}{\sin(\frac{1}{2}\hat{\omega})} e^{-j\hat{\omega}(L-1)/2}$
$r_L[n] e^{j\hat{\omega}_0 n}$	$\frac{\sin(\frac{1}{2}L(\hat{\omega} - \hat{\omega}_0))}{\sin(\frac{1}{2}(\hat{\omega} - \hat{\omega}_0))} e^{-j(\hat{\omega} - \hat{\omega}_0)(L-1)/2}$
$\frac{\sin(\hat{\omega}_b n)}{\pi n}$	$u(\hat{\omega} + \hat{\omega}_b) - u(\hat{\omega} - \hat{\omega}_b) = \begin{cases} 1 & \hat{\omega} \leq \hat{\omega}_b \\ 0 & \hat{\omega}_b < \hat{\omega} \leq \pi \end{cases}$
$a^n u[n] \quad (a < 1)$	$\frac{1}{1 - ae^{-j\hat{\omega}}}$

Table of DTFT Properties		
<i>Property Name</i>	<i>Time-Domain: $x[n]$</i>	<i>Frequency-Domain: $X(e^{j\hat{\omega}})$</i>
Periodic in $\hat{\omega}$		$X(e^{j(\hat{\omega} + 2\pi)}) = X(e^{j\hat{\omega}})$
Linearity	$ax_1[n] + bx_2[n]$	$aX_1(e^{j\hat{\omega}}) + bX_2(e^{j\hat{\omega}})$
Conjugate Symmetry	$x[n]$ is real	$X(e^{-j\hat{\omega}}) = X^*(e^{j\hat{\omega}})$
Conjugation	$x^*[n]$	$X^*(e^{-j\hat{\omega}})$
Time-Reversal	$x[-n]$	$X(e^{-j\hat{\omega}})$
Delay (n_d =integer)	$x[n - n_d]$	$e^{-j\hat{\omega}n_d} X(e^{j\hat{\omega}})$
Frequency Shift	$x[n] e^{j\hat{\omega}_0 n}$	$X(e^{j(\hat{\omega} - \hat{\omega}_0)})$
Modulation	$x[n] \cos(\hat{\omega}_0 n)$	$\frac{1}{2} X(e^{j(\hat{\omega} - \hat{\omega}_0)}) + \frac{1}{2} X(e^{j(\hat{\omega} + \hat{\omega}_0)})$
Convolution	$x[n] * h[n]$	$X(e^{j\hat{\omega}}) H(e^{j\hat{\omega}})$
Autocorrelation	$x^*[-n] * x[n]$	$ X(e^{j\hat{\omega}}) ^2$
Parseval's Theorem	$\sum_{n=-\infty}^{\infty} x[n] ^2$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\hat{\omega}}) ^2 d\hat{\omega}$