

GEORGIA INSTITUTE OF TECHNOLOGY
SCHOOL OF ELECTRICAL AND COMPUTER ENGINEERING

ECE 2026 — Summer 2024
Quiz #2

July 17, 2024

NAME: _____
(FIRST) (LAST)

GT username: _____
(e.g., gtxyz123)

Important Notes:

- Closed book, except for two double-sided pages (8.5" × 11") of hand-written notes.
- No calculators or other electronics (no smartphones/readers/watches/tablets/laptops/etc.)
- JUSTIFY your reasoning CLEARLY to receive partial credit.
- Express all angles as a fraction of π . For example, write 0.1π as opposed to 18° or 0.3142 radians.
- You must write your answer in the space provided on the exam paper itself. Only these answers will be graded. Write your answers in the provided answer boxes.
- Do not write on the backs of pages, only the fronts will be graded.

| Problem | Value | Score Earned |
|---------|-------|--------------|
| 1 | 25 | |
| 2 | 25 | |
| 3 | 25 | |
| 4 | 25 | |
| Total | | |

PROB. Su24-Q2.2. Consider a 5-point averager with input $x[n]$ and output $y[n] = \frac{1}{5} \sum_{k=0}^4 x[n-k]$.

(a) Which of the following best describes this filter? [HPF] [LPF] [BPF] [NOTCH] [ALL-PASS].

(b) The dc gain of this filter is .

(c) Specify two distinct frequencies $\hat{\omega}_1 \in [0, \pi]$ that are *nulled*, so that $y[n] = 0$ when $x[n] = \cos(\hat{\omega}_1 n)$:

$\hat{\omega}_1 =$,
 $\in [0, \pi]$
 or
 $\hat{\omega}_1 =$.
 $\in [0, \pi]$

(d) If the input is $x[n] = -5\delta[n] + 15\delta[n-2]$, the outputs at times $n \in \{0, 1, \dots, 6\}$ are:

$y[0] =$,
 $y[1] =$,
 $y[2] =$,
 $y[3] =$,
 $y[4] =$,
 $y[5] =$,
 $y[6] =$.

(e) If a sinusoidal input $x[n] = A \cos(\hat{\omega}_0 n + \varphi)$ in standard form results in the output $y[n] = \cos(0.5\pi n)$, then it must be that:

$\hat{\omega}_0 =$,
 $\in [0, \pi]$
 $A =$,
 > 0
 $\varphi =$.
 $\in (-\pi, \pi]$

PROB. Su24-Q2.3. Match each impulse response to its corresponding magnitude response by writing a letter (from A to M) in each answer box:
 (the y-axis scale is not specified, only the shapes matter)

$$h[n] = \frac{\sin(0.3\pi n)}{\pi n}$$

$$h[n] = \delta[n] - \frac{\sin(0.1\pi n)}{\pi n}$$

$$h[n] = \delta[n] - \frac{\sin(0.8\pi n)}{\pi n}$$

$$h[n] = \cos(0.6\pi n) \frac{\sin(0.1\pi n)}{\pi n}$$

$$h[n] = \cos(0.5\pi n) \frac{\sin(0.3\pi n)}{\pi n}$$

$$h[n] = \delta[n] + \frac{\sin(0.2\pi n)}{\pi n}$$

$$h[n] = \delta[n] - \cos(0.5\pi n) \frac{\sin(0.2\pi n)}{\pi n}$$

$$\frac{\sin(0.7\pi n)}{\pi n} - \cos(0.6\pi n) \frac{\sin(0.1\pi n)}{\pi n}$$

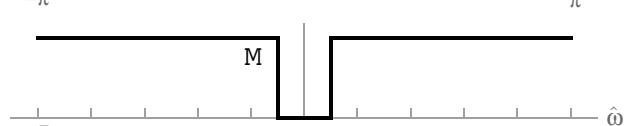
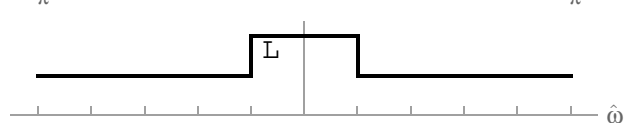
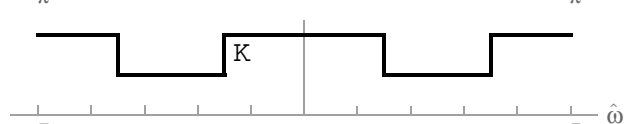
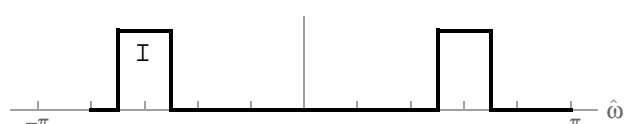
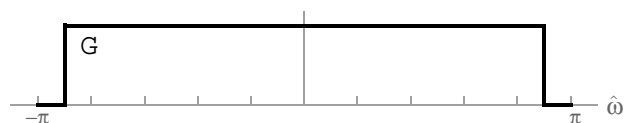
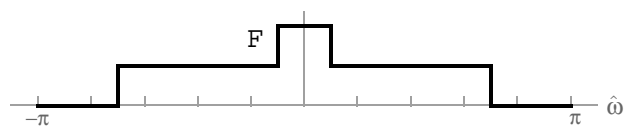
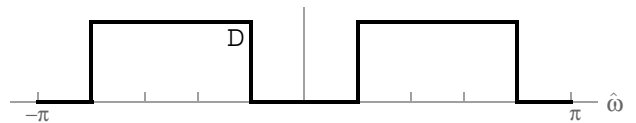
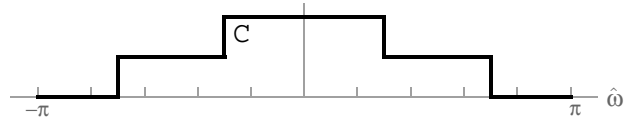
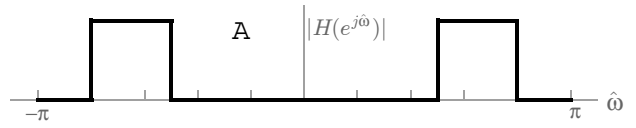
$$h[n] = \delta[n] - \cos(\pi n) \frac{\sin(0.1\pi n)}{\pi n}$$

$$\left(\frac{\sin(0.8\pi n)}{\pi n} - \frac{\sin(0.2\pi n)}{\pi n}\right) * \left(\frac{\sin(0.9\pi n)}{\pi n} - \frac{\sin(0.5\pi n)}{\pi n}\right)$$

$$h[n] = \frac{\sin(0.1\pi n)}{\pi n} + \frac{\sin(0.7\pi n)}{\pi n}$$

$$h[n] = \cos(0.2\pi n) \frac{\sin(0.5\pi n)}{\pi n}$$

$$h[n] = \delta[n] - 2\cos(0.5\pi n) \frac{\sin(0.2\pi n)}{\pi n}$$



PROB. Su24-Q2.4.

(a) The 6-point DFT of $[x[0], \dots, x[3]] = [1, 0, 0, 1]$ is:

$$X[0] = \boxed{},$$

$$X[1] = \boxed{},$$

$$X[2] = \boxed{},$$

$$X[3] = \boxed{},$$

$$X[4] = \boxed{},$$

$$X[5] = \boxed{}.$$

(b) Let a and b be unspecified real numbers.

If the 4-point DFT of $[a, b, 1]$ is $[X[0], \dots, X[3]] = [C, 2 - j, D, 2 + j]$, then:

$$a = \boxed{},$$

$$b = \boxed{},$$

$$C = \boxed{},$$

$$D = \boxed{}.$$

| Table of DTFT Pairs | |
|--|--|
| Time-Domain: $x[n]$ | Frequency-Domain: $X(e^{j\hat{\omega}})$ |
| $\delta[n]$ | 1 |
| $\delta[n - n_0]$ | $e^{-j\hat{\omega}n_0}$ |
| $r_L[n] = u[n] - u[n - L]$ | $\frac{\sin(\frac{1}{2}L\hat{\omega})}{\sin(\frac{1}{2}\hat{\omega})} e^{-j\hat{\omega}(L-1)/2}$ |
| $r_L[n]e^{j\hat{\omega}_0n}$ | $\frac{\sin(\frac{1}{2}L(\hat{\omega} - \hat{\omega}_0))}{\sin(\frac{1}{2}(\hat{\omega} - \hat{\omega}_0))} e^{-j(\hat{\omega} - \hat{\omega}_0)(L-1)/2}$ |
| $\frac{\sin(\hat{\omega}_b n)}{\pi n}$ | $u(\hat{\omega} + \hat{\omega}_b) - u(\hat{\omega} - \hat{\omega}_b) = \begin{cases} 1 & \hat{\omega} \leq \hat{\omega}_b \\ 0 & \hat{\omega}_b < \hat{\omega} \leq \pi \end{cases}$ |
| $a^n u[n] \quad (a < 1)$ | $\frac{1}{1 - ae^{-j\hat{\omega}}}$ |

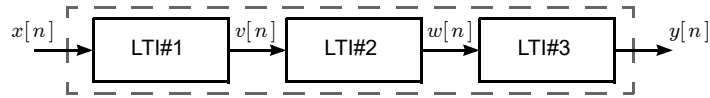
| Table of DTFT Properties | | |
|----------------------------------|--------------------------------------|---|
| Property Name | Time-Domain: $x[n]$ | Frequency-Domain: $X(e^{j\hat{\omega}})$ |
| Periodic in $\hat{\omega}$ | | $X(e^{j(\hat{\omega}+2\pi)}) = X(e^{j\hat{\omega}})$ |
| Linearity | $ax_1[n] + bx_2[n]$ | $aX_1(e^{j\hat{\omega}}) + bX_2(e^{j\hat{\omega}})$ |
| Conjugate Symmetry | $x[n]$ is real | $X(e^{-j\hat{\omega}}) = X^*(e^{j\hat{\omega}})$ |
| Conjugation | $x^*[n]$ | $X^*(e^{-j\hat{\omega}})$ |
| Time-Reversal | $x[-n]$ | $X(e^{-j\hat{\omega}})$ |
| Delay ($n_d = \text{integer}$) | $x[n - n_d]$ | $e^{-j\hat{\omega}n_d} X(e^{j\hat{\omega}})$ |
| Frequency Shift | $x[n]e^{j\hat{\omega}_0n}$ | $X(e^{j(\hat{\omega} - \hat{\omega}_0)})$ |
| Modulation | $x[n] \cos(\hat{\omega}_0n)$ | $\frac{1}{2}X(e^{j(\hat{\omega} - \hat{\omega}_0)}) + \frac{1}{2}X(e^{j(\hat{\omega} + \hat{\omega}_0)})$ |
| Convolution | $x[n] * h[n]$ | $X(e^{j\hat{\omega}})H(e^{j\hat{\omega}})$ |
| Autocorrelation | $x[-n] * x[n]$ | $ X(e^{j\hat{\omega}}) ^2$ |
| Parseval's Theorem | $\sum_{n=-\infty}^{\infty} x[n] ^2$ | $\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\hat{\omega}}) ^2 d\hat{\omega}$ |

| Table of Pairs for N-point DFT | |
|---|---|
| <i>Time-Domain: $x[n]$, $n = 0, 1, 2, \dots, N - 1$</i> | <i>Frequency-Domain: $X[k]$, $k = 0, 1, 2, \dots, N - 1$</i> |
| If $x[n]$ is finite length, i.e., $x[n] \neq 0$ only when $n \in [0, N - 1]$ and the DTFT of $x[n]$ is $X(e^{j\hat{\omega}})$ | $X[k] = X(e^{j\hat{\omega}}) \Big _{\hat{\omega}=2\pi k/N}$ (frequency sampling the DTFT) |
| $\delta[n]$ | 1 |
| 1 | $N\delta[k]$ |
| $\delta[n - n_0]$ | $e^{-j(2\pi k/N)n_0}$ |
| $e^{j(2\pi n/N)k_0}$ | $N\delta[k - k_0]$, when $k_0 \in [0, N - 1]$ |
| $r_L[n] = u[n] - u[n - L]$, when $L \leq N$ | $\frac{\sin(\frac{1}{2}L(2\pi k/N))}{\sin(\frac{1}{2}(2\pi k/N))} e^{-j(2\pi k/N)(L-1)/2}$ |
| $r_L[n]e^{j(2\pi k_0/N)n}$, when $L \leq N$ | $\frac{\sin(\frac{1}{2}L(2\pi(k - k_0)/N))}{\sin(\frac{1}{2}(2\pi(k - k_0)/N))} e^{-j(2\pi(k - k_0)/N)(L-1)/2}$ |
| $\frac{\sin(\frac{1}{2}L(2\pi n/N))}{\sin(\frac{1}{2}(2\pi n/N))} e^{j(2\pi n/N)(L-1)/2}$ | $N(u[k] - u[k - L])$, when $L \leq N$ |

| Table of DFT Properties | | |
|--------------------------------|---|---|
| <i>Property Name</i> | <i>Time-Domain: $x[n]$</i> | <i>Frequency-Domain: $X[k]$</i> |
| Periodic | $x[n] = x[n + N]$ | $X[k] = X[k + N]$ |
| Linearity | $ax_1[n] + bx_2[n]$ | $aX_1[k] + bX_2[k]$ |
| Conjugate Symmetry | $x[n]$ is real | $X[N - k] = X^*[k]$ |
| Conjugation | $x^*[n]$ | $X^*[N - k]$ |
| Time-Reversal | $x[N - n]$ | $X[N - k]$ |
| Delay (PERIODIC) | $x[n - n_d]$ | $e^{-j(2\pi k/N)n_d} X[k]$ |
| Frequency Shift | $x[n]e^{j(2\pi k_0/N)n}$ | $X[k - k_0]$ |
| Modulation | $x[n] \cos((2\pi k_0/N)n)$ | $\frac{1}{2}X[k - k_0] + \frac{1}{2}X[k + k_0]$ |
| Convolution (PERIODIC) | $x[n] * h[n] = \sum_{m=0}^{N-1} h[m]x[n - m]$ | $X[k]H[k]$ |
| Parseval's Theorem | $\sum_{n=0}^{N-1} x[n] ^2 = \frac{1}{N} \sum_{k=0}^{N-1} X[k] ^2$ | |

PROB. Su24-Q2.1.

Suppose that an “overall” system (dashed box) is constructed by cascading three FIR filters as follows:



where:

- the first system has frequency response $H_1(e^{j\hat{\omega}}) = 2 - e^{-j\hat{\omega}}$;
- the second system has impulse response $h_2[n] = 2\delta[n] + \delta[n - 1]$;
- the third system has difference equation $y[n] = \alpha w[n] + \beta w[n - 2]$, with α and β unspecified:

(a) Find the output of the *first system at time 2026* when the input is $x[n] = \cos(0.5\pi n)$:

$$\begin{aligned}
 H_1(e^{j\hat{\omega}}) &= 2 - e^{-j\hat{\omega}} \\
 \Rightarrow v[n] &= 2x[n] - x[n - 1] \\
 \Rightarrow v[2026] &= 2x[2026] - x[2025] \\
 &= -2 - 0
 \end{aligned}$$

$v[2026] =$ -2

(b) Find the *overall* output $y[n]$ when the input is a constant $x[n] = 1$ (for all n), and assuming $\alpha = \beta = 2$:

$y[n] =$ 12

$$\begin{aligned}
 \text{Overall dc gain} &= (\text{dc gain \#1})(\text{dc gain \#2})(\text{dc gain \#3}) \\
 &= (1)(3)(4) \\
 &= 12
 \end{aligned}$$

(c) If the *overall* impulse response (of the dashed box) is $h[n] = A\delta[n] - \delta[n - d]$, and if $\beta \neq 0$, then it must be that:

The impulse response of cascade of just first two is:

$$(2 \quad -1) * (2 \quad 1) = 4 \quad 0 \quad -1$$

Convolving this with $(\alpha \quad 0 \quad \beta)$ yields overall $h[n]$:

$$\begin{array}{cccccc}
 4\alpha & 0 & -\alpha & & & \\
 & & 4\beta & 0 & -\beta & \\
 \hline
 4\alpha & 0 & (-\alpha+4\beta) & 0 & -\beta & \\
 \textcircled{3} \quad A = 16 & & & & \textcircled{1} \quad \beta = 1, d = 4 & \\
 & & \textcircled{2} \quad \alpha = 4\beta = 4 & & &
 \end{array}$$

$\alpha =$ 4,

$\beta =$ 1,

$A =$ 16,

$d =$ 4.

PROB. Su24-Q2.2. Consider a 5-point averager with input $x[n]$ and output $y[n] = \frac{1}{5} \sum_{k=0}^4 x[n-k]$.

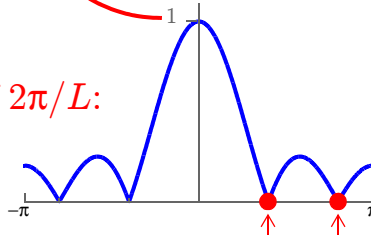
(a) Which of the following best describes this filter? [HPF] **[LPF]** [BPF] [NOTCH] [ALL-PASS].

(b) The dc gain of this filter is

1

(c) Specify two distinct frequencies $\hat{\omega}_1 \in [0, \pi]$ that are *nulled*, so that $y[n] = 0$ when $x[n] = \cos(\hat{\omega}_1 n)$:

Integer multiples of $2\pi/L$:



$\hat{\omega}_1 =$ 0.4π,
 $\in [0, \pi]$

or

$\hat{\omega}_1 =$ 0.8π,
 $\in [0, \pi]$

(d) If the input is $x[n] = -5\delta[n] + 15\delta[n-2]$, the outputs at times $n \in \{0, 1, \dots, 6\}$ are:

$y[0] =$ -1,

$y[1] =$ -1,

$y[2] =$ 2,

$y[3] =$ 2,

$y[4] =$ 2,

$y[5] =$ 3,

$y[6] =$ 3.

Convolve 0.2 0.2 0.2 0.2 0.2 with -5 0 15:

$$\begin{array}{cccccc}
 -1 & -1 & -1 & -1 & -1 & & \\
 & & 3 & 3 & 3 & 3 & 3 \\
 \hline
 -1 & -1 & 2 & 2 & 2 & 3 & 3
 \end{array}$$

(e) If a sinusoidal input $x[n] = A \cos(\hat{\omega}_0 n + \phi)$ in standard form results in the output $y[n] = \cos(0.5\pi n)$, then it must be that:

When $\hat{\omega} = 0.5\pi$, the frequency response becomes a polynomial in $e^{-j0.5\pi} = -j$:

$$\Rightarrow H(e^{j0.5\pi}) = 0.2(1 - j + j^2 - j^3 + j^4) = 0.2$$

\Rightarrow No phase change, attenuates by factor of 0.2

$\hat{\omega}_0 =$ 0.5π,
 $\in [0, \pi]$

$A =$ 5,
 > 0

$\phi =$ 0,
 $\in (-\pi, \pi]$

PROB. Su24-Q2.3. Match each impulse response to its corresponding magnitude response by writing a letter (from A to M) in each answer box:
 (the y-axis scale is not specified, only the shapes matter)

J $h[n] = \frac{\sin(0.3\pi n)}{\pi n}$

M $h[n] = \delta[n] - \frac{\sin(0.1\pi n)}{\pi n}$

E $h[n] = \delta[n] - \frac{\sin(0.8\pi n)}{\pi n}$

I $h[n] = \cos(0.6\pi n) \frac{\sin(0.1\pi n)}{\pi n}$

D $h[n] = \cos(0.5\pi n) \frac{\sin(0.3\pi n)}{\pi n}$

L $h[n] = \delta[n] + \frac{\sin(0.2\pi n)}{\pi n}$

K $h[n] = \delta[n] - \cos(0.5\pi n) \frac{\sin(0.2\pi n)}{\pi n}$

H $\frac{\sin(0.7\pi n)}{\pi n} - \cos(0.6\pi n) \frac{\sin(0.1\pi n)}{\pi n}$

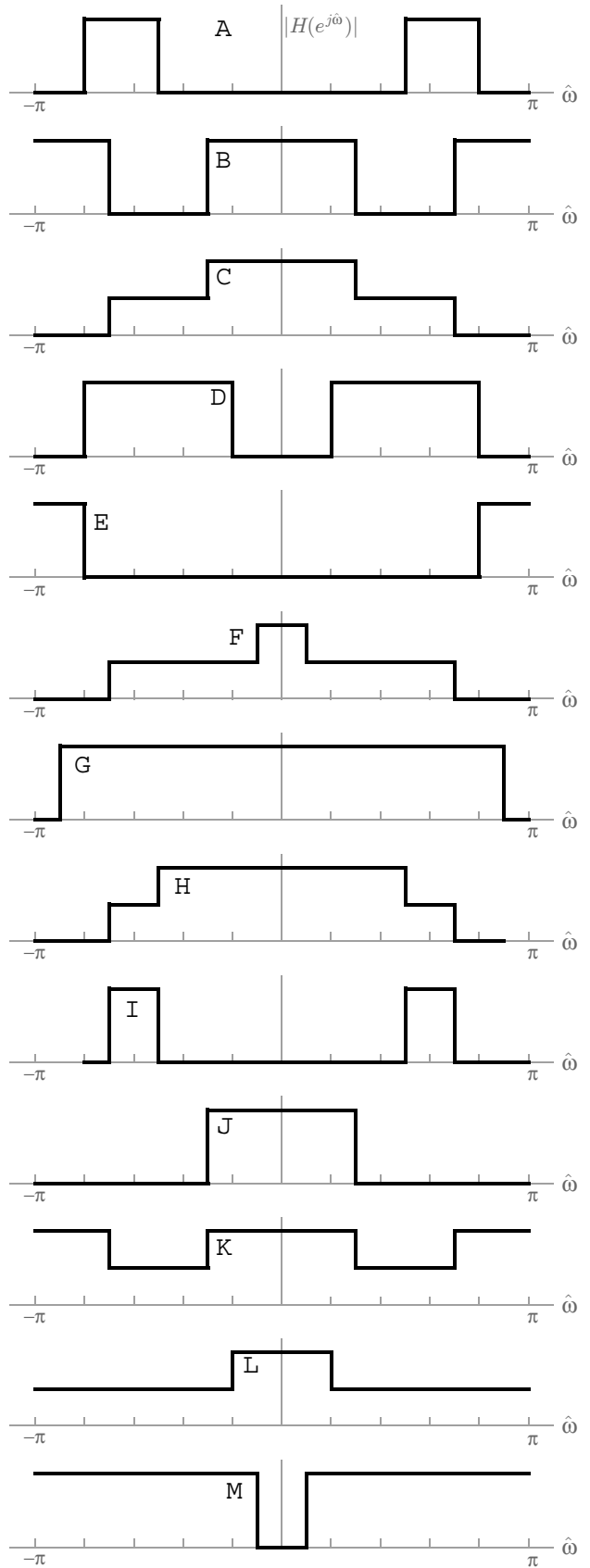
G $h[n] = \delta[n] - \cos(\pi n) \frac{\sin(0.1\pi n)}{\pi n}$

A $\left(\frac{\sin(0.8\pi n)}{\pi n} - \frac{\sin(0.2\pi n)}{\pi n}\right) * \left(\frac{\sin(0.9\pi n)}{\pi n} - \frac{\sin(0.5\pi n)}{\pi n}\right)$

F $h[n] = \frac{\sin(0.1\pi n)}{\pi n} + \frac{\sin(0.7\pi n)}{\pi n}$

C $h[n] = \cos(0.2\pi n) \frac{\sin(0.5\pi n)}{\pi n}$

B $h[n] = \delta[n] - 2\cos(0.5\pi n) \frac{\sin(0.2\pi n)}{\pi n}$



PROB. Su24-Q2.4.

(a) The 6-point DFT of $[x[0], \dots, x[3]] = [1, 0, 0, 1]$ is:

The DTFT is $X(e^{j\hat{\omega}}) = 1 + e^{-j3\hat{\omega}}$

\Rightarrow sampling at $k(2\pi/6) = k\pi/3$ yields:

$$X[k] = 1 + e^{-jk\pi}$$

$$= \begin{cases} 2 & \text{when } k \text{ is even,} \\ 0 & \text{when } k \text{ is odd.} \end{cases}$$

$$\begin{aligned} X[0] &= \boxed{2}, \\ X[1] &= \boxed{0}, \\ X[2] &= \boxed{2}, \\ X[3] &= \boxed{0}, \\ X[4] &= \boxed{2}, \\ X[5] &= \boxed{0}. \end{aligned}$$

(b) Let a and b be unspecified real numbers.

If the 4-point DFT of $[a, b, 1]$ is $[X[0], \dots, X[3]] = [C, 2 - j, D, 2 + j]$, then:

Sampling $X(e^{j\hat{\omega}}) = a + be^{-j\hat{\omega}} + e^{-j2\hat{\omega}}$ at $2\pi/4 = 0.5\pi$ yields

$$X[1] = a - jb - 1.$$

Equate with given $2 - j \Rightarrow a = 3, b = 1$

$$\begin{aligned} a &= \boxed{3}, \\ b &= \boxed{1}, \\ C &= \boxed{5}, \\ D &= \boxed{3}. \end{aligned}$$

Similarly, sampling at $0(0.5\pi)$ and $2(0.5\pi)$ yields:

$$C = X[0] = a + b + 1 = 5,$$

$$D = X[2] = a - b + 1 = 3.$$