## GEORGIA INSTITUTE OF TECHNOLOGY SCHOOL OF ELECTRICAL AND COMPUTER ENGINEERING

ECE 2026 - Summer 2014
Quiz \#2
July 9, 2014


Circle your recitation section (otherwise you lose 3 points!):

|  | Mon | Tue |
| :---: | :---: | :---: |
| $\begin{array}{r} 10-11: 45 \mathrm{am} \\ 12-1: 45 \mathrm{pm} \end{array}$ |  | L02 (Moore) |
|  |  | L03 (Moore) |
| 2:40-3:50pm |  | L04 (Davis) |
| 4-5:45pm | L01 (Barry) |  |

## Important Notes:

- Circle your recitation above.
- DO NOT unstaple the test.
- One two-sided page ( $8.5 " \times 11$ ") of hand-written notes permitted.
- JUSTIFY your reasoning CLEARLY to receive partial credit.
- You must write your answer in the space provided on the exam paper itself. Only these answers will be graded. Circle your answers, or write them in the boxes provided. If more space is needed for scratch work, use the backs of the previous pages.

| Problem | Value | Score Earned |  |
| :---: | :---: | :---: | :---: |
| 1 | 20 |  |  |
| 2 | 20 |  |  |
| 3 | 20 |  |  |
| 4 | 20 |  |  |
| 5 | 20 |  |  |
| No/Wrong Rec | -3 |  |  |
| Total |  |  |  |
|  |  |  |  |

## PROB. Su14-Q2.1.

(a) The acronym FIR stands for $\square$ finite impulse response
(b) The acronym LTI stands for linear (and) time-invariant
(c) The acronym DTFT stands for discrete-time Fourier transform
(d) The acronym $D F T$ stands for
discrete Fourier transform
(e) The acronym FFT stands for

## fast Fourier transform

(f) Convolving a length- 5 signal with a length-10 signal yields a length-
(g) The response of an LTI system with frequency response $H\left(e^{j \omega}\right)=20 \cos (4 \hat{\omega})$ to a sinusoidal input of

$$
x[n]=\sin (\pi n / 12) \quad \text { is } \quad y[n]=\quad 10 \sin (\pi n / 12)
$$

$$
@ \hat{\omega}=\pi / 12, H\left(e^{j \hat{\omega}}\right)=20 \cos (\pi / 30)=10
$$

(h) There are $6 \quad \begin{array}{r}\text { values of } \hat{\omega} \text { in the range } 0<\hat{\omega}<\pi \text { for which the frequency response } H\left(e^{j \hat{\omega}}\right) \\ \left.\text { of a } 13 \text {-point averager is zero (i.e., for which } H\left(e^{j \hat{\omega}}\right)=0\right) .\end{array}$ $H\left(e^{j \hat{\omega}}\right)=e^{-j 6 \hat{\omega}} \frac{\sin (13 \omega / 2)}{13 \sin (\hat{\omega} / 2)}$ is zero when numerator is zero; i.e. when $13 \hat{\omega} / 2=k \pi$ for $k=1,2, \ldots 6$. (Other values of $k$ fall outside the 0 -to- $\pi$ range.)
(i) The inverse DTFT of $X\left(e^{j \hat{\omega}}\right)=\frac{e^{-j \omega}}{1-0.8 e^{-j \omega}}$ is $x[n]=0.8^{n-1} u[n-1]$.
(Give an equation that is valid for all $n$.)
(j) $\square$ TRUE FALSE If $x[n]$ is real then its $\operatorname{DFT} X[k]$ is also real.
(k) After taking the 16 -point DFT of $[x[0], x[1], x[2], x[3]]=[2,4,6,8]$, the zero-th DFT coefficient is:

$$
X[0]=20
$$

PROB. Su14-Q2.2. (The three parts of this problem are unrelated to eachother.)
(a) In the space below, sketch the frequency response of an LTI system whose impulse response is

(b) Write an equation for the frequency response $H\left(e^{j \hat{\omega}}\right)$ of an LTI system when the impulse response $h[n]$ is shown in the stem plot below:

$$
\begin{aligned}
& \Rightarrow H\left(e^{j \hat{\omega}}\right)=4+3 e^{-5 j \hat{\omega}}+e^{-6 j \hat{\omega}}
\end{aligned}
$$

(c) Consider an LTI system whose frequency response is $H\left(e^{j \hat{\omega}}\right)=e^{-j 3 \hat{\omega}}\left|H\left(e^{j \hat{\omega}}\right)\right|$, where $\left|H\left(e^{j \omega}\right)\right|$ oscillates sinusoidally between 1 and 4 , with a period of $2 \pi / 3$, as shown below:

$$
\begin{aligned}
& \Rightarrow H\left(e^{j \hat{\omega}}\right)=e^{-j 3 \hat{\omega}}(2.5+1.5 \cos (3 \hat{\omega})) \quad \quad{ }^{\wedge}\left|H\left(e^{j \hat{\omega}}\right)\right|=2.5+1.5 \cos (3 \hat{\omega}) \\
& =e^{-j 3 \hat{\omega}}\left(2.5+0.75 e^{j 3 \hat{\omega}}+0.75 e^{-j 3 \hat{\omega}}\right) \\
& =\underline{0.75}+\underline{2.5} e^{-j 3 \hat{\omega}}+\underline{0.75} e^{-j 6 \hat{\omega}}
\end{aligned}
$$

Find numerical values for the impulse response $h[n]$ in time domain, at times $n \in\{0,1,2, \ldots 6\}$ :

| $h[0]$ | $h[1]$ | $h[2]$ | $h[3]$ | $h[4]$ | $h[5$ | $h[6]$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0.75 | 0 | 0 | 2.5 | 0 | 0 | 0.75 |

PROB. Su14-Q2.3. The figure depicts a cascade connection of two LTI systems, where the output $w[n]$ of system\#1 is the input to system\#2, and the output of system\#2 is the overall output:


LTI system\#1 is defined by the difference equation $w[n]=x[n]+2 x[n-2]+x[n-4]$.
LTI system\#2 is defined by the impulse response $\quad h_{2}[n]=\delta[n]+\delta[n-1]$.
(a) The DC gain of system\#1 is $\square$ since $H_{1}\left(e^{j 0}\right)=1+2+1=4$.
(b) In response to $x[n]=\cos \left(\hat{\omega}_{1} n\right)$, the output of system\#1 will be zero for all time (i.e., $w[n]=0$ ) when the input frequency is:

$$
\begin{aligned}
& \hat{\omega}_{1}=0.5 \pi \quad \text { radians } . \\
& H_{1}\left(e^{j \hat{\omega}}\right)=1+2 e^{-2 j \hat{\omega}}+e^{-4 j \hat{\omega}} \\
& =e^{-2 j \hat{\omega}}\left(e^{2 j \hat{\omega}}+2+e^{-2 j \hat{\omega}}\right) \\
& =e^{-2 j \hat{\omega}}(2+2 \cos (2 \hat{\omega})) \\
& =0 \\
& \ldots \text { when } \hat{\omega}=0.5 \pi
\end{aligned}
$$

(c) The difference equation for the overall cascade system (relating $x[n]$ to $y[n]$ ) can be written as:

$$
y[n]=b_{0} x[n]+b_{1} x[n-1]+b_{2} x[n-2]+b_{3} x[n-3]+b_{4} x[n-4]+b_{5} x[n-5]+b_{6} x[n-6],
$$

where the coefficients are:


$$
\text { convolving } \mathbf{h}_{1}=\left[\begin{array}{lllll}
1 & 0 & 2 & 0 & 1
\end{array}\right] \text { with } \mathbf{h}_{2}=\left[\begin{array}{ll}
1 & 1
\end{array}\right] \text { yields } \mathbf{h}=\left[\begin{array}{lllll}
1 & 1 & 2 & 2 & 1
\end{array}\right]
$$

PROB. Su14-Q2.4. We have seen that LTI systems can be represented in a variety of ways.
For each of the LTI systems below, provide the missing representations. (The frequency response can be given in any convenient form; you need not simplify.)

| (a) difference equation: | $y[n]=x[n]+x[n-1]$ |
| :---: | :---: |
| impulse response: | $h[n]=\delta[n]+\delta[n-1]$ |
| frequency response: | $H\left(e^{j \hat{\omega}}\right)=1+e^{-j \hat{\omega}}$ |

extra factor of $3 e^{-j 2 \hat{\omega}}$ beyond 3 -pt averager $\Rightarrow$ scale by 3 and delay by 2
(b) frequency response:
difference equation:

|  |  |
| ---: | :--- |
| $H\left(e^{j \hat{\omega}}\right)$ | $=\left(\frac{\sin (1.5 \hat{\omega})}{\sin (0.5 \hat{\omega})}\right) e^{-j 3 \hat{\omega}}$ |
| extra factor of $3 e^{-j 2 \hat{\omega}}$ beyond 3-pt averager <br> $\Rightarrow$ scale by 3 and delay by 2 |  |
| $h[n]$ | $=x[n-2]+x[n-3]+x[n-4]$ |
|  | $=u[n-2]-u[n-5]$ |

(c) MATLAB code:

$$
\text { yy }=\operatorname{conv}(x x,[1,0,0,0,-1]) ;
$$

impulse response:

$$
h[n]=\delta[n]-\delta[n-4]
$$

frequency response:

$$
H\left(e^{j \hat{\omega}}\right)=1-e^{-j 4 \hat{\omega}}
$$

PROB. Su14-Q2.5. Find the 10 -point $\operatorname{DFT}\{X[0], X[1], \ldots X[9]\}$ of the length-10 signal segment:

$$
\begin{aligned}
& x[0]=1 \\
& x[1]=-1 \\
& x[2]=1 \\
& x[3]=-1 \\
& x[4]=1 \\
& x[5]=-1 \\
& x[6]=1 \\
& x[7]=-1 \\
& x[8]=1 \\
& x[9]=-1
\end{aligned}
$$

which is the same as prescribed by the inverse $10-$ pt DFT equation:

$$
\mathrm{x}[n]=\frac{1}{10} \sum_{k=0}^{9} X[k] e^{j k n 2 \pi / 10}
$$

when only $X[5]=10$ is nonzero.

| $X[0]=$ | 0 |
| :---: | :---: |
| $X[1]=$ | 0 |
| $X[2]=$ | 0 |
| $X[3]=$ | 0 |
| $X[4]=$ | 0 |
| $X[5]=$ | 10 |
| $X[6]=$ | 0 |
| $X[7]=$ | 0 |
| $X[8]=$ | 0 |
| $X[9]=$ | 0 |

Hint: Answers are all real integers.

