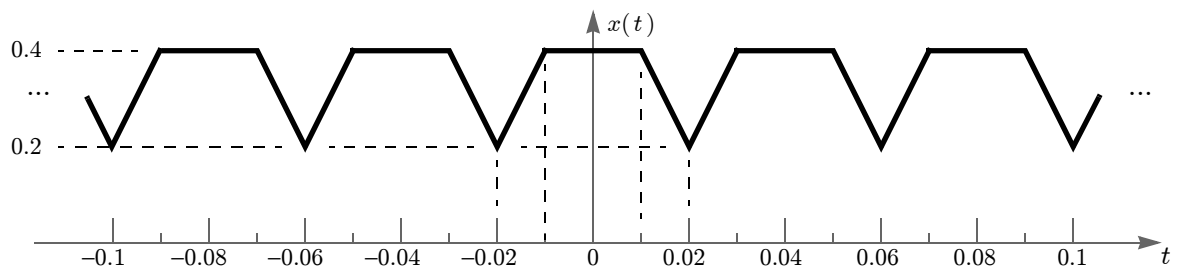
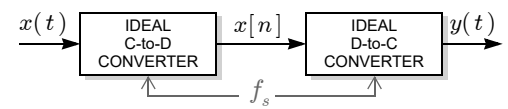


Problem	Value	Score Earned
1	35	
2	30	
3	35	
Total		

PROB. Sp25-Q2.1. Consider the periodic signal $x(t)$ shown below:



Suppose this signal is input to the cascade of an ideal C-to-D and ideal D-C converter, both with the same (unspecified) sampling rate f_s , as shown here:



- (a) In the Fourier series representation $x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j2\pi k f_0 t}$, the fundamental frequency and zero-th coefficient are:

$f_0 =$ Hz,
 ≥ 0

$a_0 =$.

- (b) If the sampling rate is $f_s = 12.5$ Hz, the D-C output can be written as $y(t) = B_1 + A_1 \cos(25\pi t)$ where:

$A_1 =$,
 ≥ 0

$B_1 =$.

- (c) If the sampling rate is $f_s = 50$ Hz, the D-C output can be written as $y(t) = B + A \cos(2\pi f_0 t + \phi)$ where:

$f_0 =$ Hz,
 ≥ 0

$\phi =$ rads
 $\in (-\pi, \pi]$

$A =$,
 ≥ 0

$B =$.

PROB. Sp25-Q2.2.

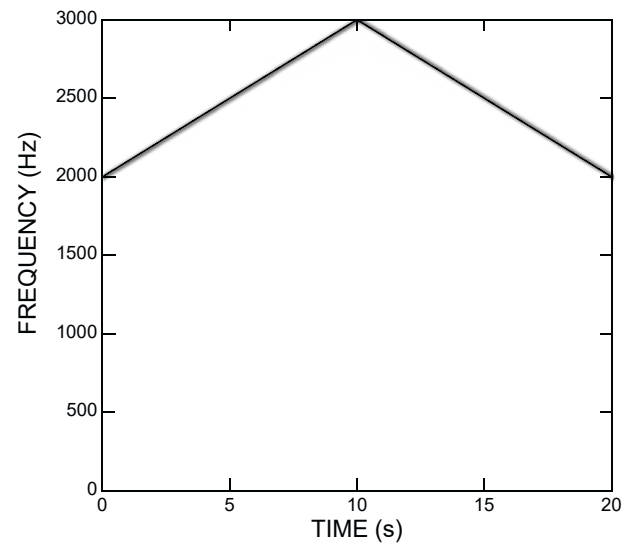
If running the shown code leads to the shown plot:

- (a) then the unspecified variables `fsamp` and `t2` are:

`fsamp` = ,

`t2` = .

```
A = 
B = 
fsamp = 
t2 = 
t = 0:1/fsamp:t2;
x = sin(A*pi*(t + B).^2);
spectrogram(x,200,[],1e5,fsamp,'yaxis')
```



- (b) Specify one possible set of numerical values for the unspecified variables `A` and `B`, with the constraint that both are positive, satisfying: $0 < A < 300$ and $0 < B < 300$.

`A` = ,

`B` = .

- (c) Specify a second possible set of numerical values for the unspecified variables `A` and `B`, this time with the constraint that both are *negative*, satisfying: $-300 < A < 0$ and $-300 < B < 0$.

`A` = ,

`B` = .

PROB. Sp25-Q2.3. All parts of this problem relate to an FIR filter whose impulse response has the form:

$$h[n] = 3\delta[n] + 5\delta[n - 1] + \beta\delta[n - 2],$$

where the value of the unspecified parameter β may be different in each part.

- (a) If the filter output at time 3 is $y[3] = 3$ when the filter input is $x[n] = \frac{2}{1 + n^2}$, then $\beta =$.

- (b) Let $s[n]$ denote the *step response* of the filter (i.e., the filter output in response to the unit step input $u[n]$, where $u[n] = 0$ for $n < 0$ and $u[n] = 1$ for $n \geq 0$). If $s[n] = 0$ for all $n \notin \{0, 1, 2\}$, then it must be that:

$$\beta =$$
 ,

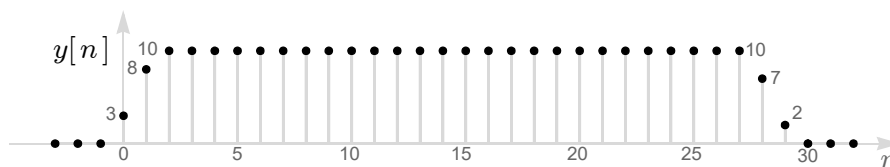
$$s[0] =$$
 ,

$$s[1] =$$
 ,

$$s[2] =$$
 .

- (c) Let $z[n] = h[n] * h[n]$ denote the convolution of $h[n]$ with itself. If $z[2] = 27$ then $\beta =$.

- (d) If the rectangular filter input $x[n] = u[n] - u[n - L]$ results in the filter output shown below, then:

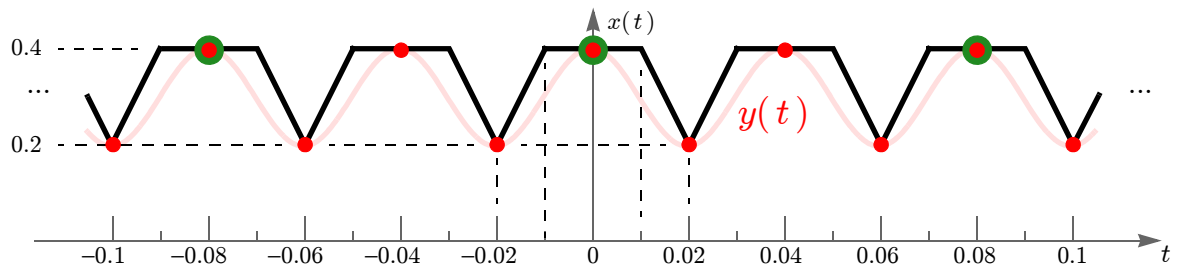


$$\beta =$$
 ,

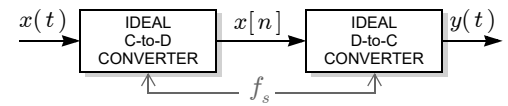
$$L =$$
 .

Problem	Value	Score Earned
1	35	
2	30	
3	35	
Total		

PROB. Sp25-Q2.1. Consider the periodic signal $x(t)$ shown below:



Suppose this signal is input to the cascade of an ideal C-to-D and ideal D-C converter, both with the same (unspecified) sampling rate f_s , as shown here:



- (a) In the Fourier series representation $x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j2\pi k f_0 t}$, the fundamental frequency and zero-th coefficient are:

$$\text{Measure } T_0 = 0.04 \Rightarrow f_0 = \frac{1}{T_0} = 25 \text{ Hz}$$

$$a_0 = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) dt = 25(0.01)(0.3 + 0.4 + 0.4 + 0.3) = 0.35$$

$$f_0 = 25 \text{ Hz}, \quad a_0 = 0.35$$

- (b) If the sampling rate is $f_s = 12.5$ Hz, the D-C output can be written as $y(t) = B_1 + A_1 \cos(25\pi t)$ where:

$$x[n] = x\left(\frac{n}{f_s}\right) = x(0.08n) = 0.4 \text{ (constant) for all } n$$

(see green dots in figure)

$$\Rightarrow y(t) = 0.4 \text{ (constant) for all } t$$

$$A_1 = 0, \quad B_1 = 0.4$$

- (c) If the sampling rate is $f_s = 50$ Hz, the D-C output can be written as $y(t) = B + A \cos(2\pi f_0 t + \phi)$ where:

$$x[n] = x\left(\frac{n}{f_s}\right) = x(0.02n) = \text{red dots in the figure}$$

$$= 0.3 + 0.1 \cos(\pi n)$$

$$\Rightarrow y(t) = x[n]|_{n=f_s t}$$

$$= 0.3 + 0.1 \cos(50\pi t)$$

$$= \text{pink curve in figure}$$

$$f_0 = 25 \text{ Hz}, \quad \phi = 0 \text{ rad}, \quad A = 0.1, \quad B = 0.3$$

PROB. Sp25-Q2.2.

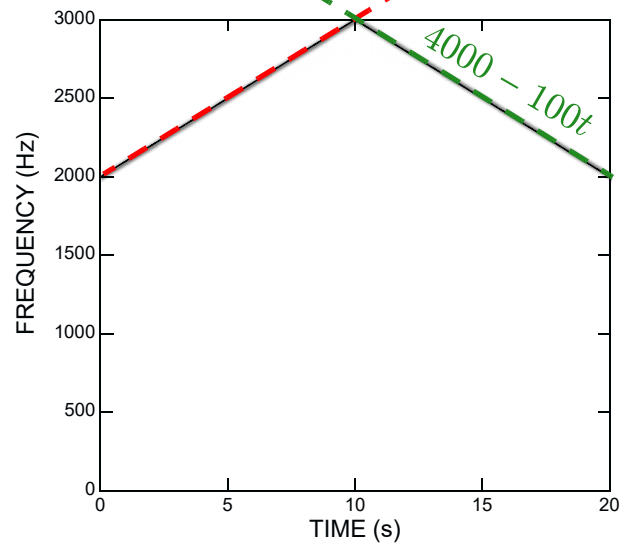
If running the shown code leads to the shown plot:

- (a) then the unspecified variables fsamp and t2 are:

fsamp = ,
> 0

t2 = .
> 0

```
A = 
B = 
fsamp = 
t2 = 
t = 0:1/fsamp:t2;
x = sin(A*pi*(t + B).^2);
spectrogram(x,200,[],1e5,fsamp,'yaxis')
```



- (b) Specify one possible set of numerical values for the unspecified variables A and B, with the constraint that both are positive, satisfying: $0 < A < 300$ and $0 < B < 300$.

$$f_i(t) = \frac{1}{2\pi} \frac{d}{dt} (A\pi(t + B)^2)$$

$$= AB + At$$

Equate this to equation for red dashed line,
assuming aliasing happens after $t > 10$:

$$= 2000 + 100t$$

$$\Rightarrow A = 100$$

$$\Rightarrow B = 20$$

A = ,
> 0

B = .
> 0

- (c) Specify a second possible set of numerical values for the unspecified variables A and B, this time with the constraint that both are *negative*, satisfying: $-300 < A < 0$ and $-300 < B < 0$.

Equate

$$f_i(t) = AB + At$$

to equation for green dashed line,
assuming aliasing happens before $t < 10$:

$$= 4000 - 100t$$

$$\Rightarrow A = -100$$

$$\Rightarrow B = -40$$

A = ,
< 0

B = .
< 0

PROB. Sp25-Q2.3. All parts of this problem relate to an FIR filter whose impulse response has the form:

$$h[n] = 3\delta[n] + 5\delta[n-1] + \beta\delta[n-2],$$

where the value of the unspecified parameter β may be different in each part.

- (a) If the filter output at time 3 is $y[3] = 3$ when the filter input is $x[n] = \frac{2}{1+n^2}$, then $\beta =$ 0.4.

$$\begin{aligned} y[3] &= 3x[3] + 5x[2] + \beta x[1] \\ &= 3\left(\frac{2}{1+9}\right) + 5\left(\frac{2}{1+4}\right) + \beta\left(\frac{2}{1+1}\right) \\ &= 2.6 + \beta \end{aligned}$$

- (b) Let $s[n]$ denote the *step response* of the filter (i.e., the filter output in response to the unit step input $u[n]$, where $u[n] = 0$ for $n < 0$ and $u[n] = 1$ for $n \geq 0$). If $s[n] = 0$ for all $n \notin \{0, 1, 2\}$, then it must be that:

If input is unit step, output at time 3 is:

$$\begin{aligned} s[3] &= 3u[3] + 5u[2] + \beta u[1] \\ &= 3(1) + 5(1) + \beta(1) \\ &= 8 + \beta \Rightarrow \text{zero when } \beta = -8 \end{aligned}$$

$$\Rightarrow s[0] = 3u[0] + 5u[-1] + \beta u[-2] = 3 + 0 + 0 = 3$$

$$s[1] = 3u[1] + 5u[0] + \beta u[-1] = 3 + 5 + 0 = 8$$

$$s[2] = 3u[2] + 5u[1] + \beta u[0] = 3 + 5 - 8 = 0$$

$$\beta =$$
 -8

$$s[0] =$$
 3

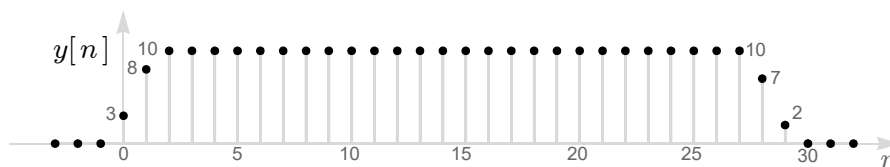
$$s[1] =$$
 8

$$s[2] =$$
 0

- (c) Let $z[n] = h[n] * h[n]$ denote the convolution of $h[n]$ with itself. If $z[2] = 27$ then $\beta =$ $\frac{1}{3}$.

$$\begin{aligned} z[2] &= 3h[2] + 5h[1] + \beta h[0] \\ &= 3\beta + 5(5) + \beta(3) \\ &= 6\beta + 25 \end{aligned}$$

- (d) If the rectangular filter input $x[n] = u[n] - u[n-L]$ results in the filter output shown below, then:



$$\beta =$$
 2

$$L =$$
 28

$$y[2] = 3u[2] + 5u[1] + \beta u[0] = 3 + 5 + \beta = 10 \Rightarrow \beta = 2$$

$$\text{Output length} = \text{input length} + \text{filter order} \Rightarrow 30 = L + 2$$