April 28, 2023

NAME:


GT username: $\qquad$ (e.g., gtxyz123)

Circle your recitation section:

| L01 (Chen) | L07 (Davenport) | L09 (Hessler) | L11 (Hessler) |
| :--- | :--- | :--- | :--- | :--- |
| L02 (Duan) | L08 (Duan) | L10 (Chen) |  |

## Important Notes:

- You may tear off the tables from the back of the exam, but otherwise do not unstaple the rest.
- One two-sided page ( 8.5 " $\times 11$ ") of notes permitted.
- Calculators are allowed, but no other electronics (no smartphones/watches/readers/tablets/etc).
- JUSTIFY your reasoning CLEARLY to receive partial credit.
- Express all angles as a fraction of $\pi$. For example, write $0.1 \pi$ as opposed to $18^{\circ}$ or 0.3142 radians.
- You must write your answer in the space provided on the exam paper itself. Only these answers will be graded. Write your answers in the provided answer boxes. If more space is needed for scratch work, use the backs of the previous pages.

| Problem | Value | Score |
| :---: | :---: | :---: |
| 1 | 15 |  |
| 2 | 15 |  |
| 3 | 18 |  |
| 4 | 10 |  |
| 5 | 12 |  |
| 7 | Total |  |
| 7 |  |  |

## PROB. Sp23-F.1.

Suppose that the following equation is true, for all time $t$ :

$$
t_{0} \cos \left(2 \pi f_{0}\left(t-t_{0}\right)\right)=\sqrt{2} \cos \left(2 \pi f_{0} t\right)+\sqrt{2} \sin \left(2 \pi f_{0} t\right) .
$$

(a) As a function of the unspecified parameters $f_{0}$ and $t_{0}$, the corresponding phasor equation is:

(b) Find the smallest positive values for $f_{0}$ and $t_{0}$ so that the given equation is true, for all time $t$ :


PROB. Sp23-F.2. $\quad$ Let $x[n]=-2 \delta[n]+2 \delta[n-2]+\delta[n-4]$ be the input to an LTI filter whose frequency response is $H\left(e^{j \hat{\omega}}\right)=e^{-j \hat{\omega}}(3+4 \cos (\hat{\omega}))$, resulting in an output $y[n]$.

(a) $\square$ (True or False): The system is FIR.
(b) The impulse response $h[n]$ satisfies:

$h[2]=\square$.
(c) The filter output $y[n]$ satisfies:


## PROB. Sp23-F.3.

Suppose that a continuous-time signal $x(t)=8 \cos (2000 \pi t) \sin (4000 \pi t)$ is fed to an ideal sampling/ filtering/reconstruction system, producing an output $y(t)$, as shown below:


The difference equation for the FIR filter is $y[n]=x[n]+B x[n-1]+x[n-2]$.
The parameter $B$ and the sampling rate $f_{s}$ are unspecified and different in each part below.
(a) How big must $f_{s}$ be to avoid aliasing?
(b) If a sampling rate of $f_{s}=12000 \mathrm{~Hz}$ results in a D-to-C output $y(t)$ that is periodic with fundamental frequency $f_{0}=3000 \mathrm{~Hz}$, then it must be that:
 $\square$. $B=\square$
(c) When $f_{s}=6000 \mathrm{~Hz}$ and $B=2$, the D-to-C output has the standard form $y(t)=A \cos \left(2 \pi f_{0} t+\varphi\right)$, where:




PROB. Sp23-F.4. Shown below are ten different impulse responses $h[n]$, labeled A through J, where $h[n]$ is plotted for times $n \in\{-5, \ldots, 40\}$ only:

(The y-axis scales are not specified, they are not needed to solve the problem.)
Each $h[n]$ shown above is the impulse response of a filter whose frequency response $H\left(e^{j \hat{\omega}}\right)$ is listed below, but the order is scrambled. Match each $H\left(e^{j \hat{\omega}}\right)$ below to its corresponding stem plot for $h[n]$ above. Indicate your answers by writing a letter (from A through J) into each of the ten answer boxes.


$$
\begin{equation*}
H_{1}\left(e^{j \hat{\omega}}\right)=\frac{e^{-j 10 \hat{\omega}}}{1-0.9 e^{-j \hat{\omega}}} \tag{i}
\end{equation*}
$$

(ii)


$$
H_{2}\left(e^{j \hat{\omega}}\right)=\frac{e^{-j 10 \hat{\omega}}}{1+0.9 e^{-j \hat{\omega}}}
$$

(iii)


$$
H_{3}\left(e^{j \hat{\omega}}\right)=\frac{1+0.9 e^{-j 10 \hat{\omega}}}{1-0.9 e^{-j \hat{\omega}}}
$$

(iv)


$$
H_{4}\left(e^{j \hat{\omega}}\right)=\frac{1-0.9 e^{-j 10 \hat{\omega}}}{1-0.9 e^{-j \hat{\omega}}}
$$

(v)


$$
H_{5}\left(e^{j \hat{\omega}}\right)=2 e^{-j 7 \hat{\omega}} \frac{\sin (2.5 \hat{\omega})}{\sin (0.5 \hat{\omega})} \cos (5 \hat{\omega})
$$



$$
\begin{equation*}
H_{6}\left(e^{j \hat{\omega}}\right)=2 j e^{-j 7 \hat{\omega}} \frac{\sin (2.5 \hat{\omega})}{\sin (0.5 \hat{\omega})} \sin (5 \hat{\omega}) \tag{vi}
\end{equation*}
$$



$$
\begin{equation*}
H_{7}\left(e^{j \hat{\omega}}\right)=\frac{0.5}{1-0.9 e^{-j(\hat{\omega}-0.1 \pi)}}+\frac{0.5}{1-0.9 e^{-j(\hat{\omega}+0.1 \pi)}} \tag{vii}
\end{equation*}
$$

(viii)


$$
H_{8}\left(e^{j \hat{\omega}}\right)=\frac{0.5}{1-0.9 e^{-j(\hat{\omega}-0.05 \pi)}}+\frac{0.5}{1-0.9 e^{-j(\hat{\omega}+0.05 \pi)}}
$$

(ix) $\square$

$$
H_{9}\left(e^{j \hat{\omega}}\right)=e^{-j 12 \hat{\omega}} \frac{\sin (2.5 \hat{\omega})}{\sin (0.5 \hat{\omega})}
$$

(x) $\square$

$$
H_{10}\left(e^{j \hat{\omega}}\right)=e^{-j 14.5 \hat{\omega}} \frac{\sin (5 \hat{\omega})}{\sin (0.5 \hat{\omega})}
$$

## PROB. Sp23-F.5.

Consider the following serial cascade of three LTI systems:


As shown above, an input sequence $x[n]$ is fed to a first LTI system, whose output $w[n]$ is fed as an input to a second system, whose output is fed to a third system, producing an overall output $y[n]$.

- The first system (with input $x[n]$ and output $w[n]$ ) has the pole-zero plot for its system function shown above, with a zero at 0.8 and a pole at -0.8 .
- The second system has frequency response: $H_{2}\left(e^{j \hat{\omega}}\right)=b_{0}+b_{1} e^{-j \hat{\omega}}+b_{2} e^{-2 j \hat{\omega}}$, where $\left\{b_{k}\right\}$ are unspecified.
- The third system has impulse response: $\quad h_{3}[n]=(0.8)^{n} u[n]$.
(a)
 (True or False): The first system is FIR.
(b)
 (True or False): The second system is FIR.
(c)
 (True or False): The third system is FIR.
(d) If the difference equation relating the overall output to the overall input is:

$$
y[n]=x[n]-x[n-1],
$$

then it must be that:

$$
b_{0}=\square,
$$

$$
b_{1}=\square
$$

$$
b_{2}=\square
$$

PROB. Sp23-F.6. $\quad$ Let $x[n]=\delta[n]+\beta \delta[n-1]+\delta[n-2]$.
Shown below are plots of $\left|\sum_{n=0}^{2} x[n] e^{-j k 2 \pi n / N}\right|$ versus $k \in\{0, \ldots N-1\}$, labeled A through L . (The axes are not labeled, only the shapes matter.)
Match each plot to the corresponding values of the parameters $\beta$ and $N$.
Indicate your answer by writing a letter (from A through $L$ ) in each answer box.
(a)

$\beta=0, N=12$.
(b)

$\beta=1, N=12$.
(c)

$\beta=2, N=12$.
(d)

$\beta=3, N=12$.
(e)

$\beta=0, N=15$.

$\beta=1, N=15$.
(g)

$\beta=2, N=15$.
(h)

$\beta=3, N=15$.
(i)

$\beta=0, N=16$.
(j)

$\beta=1, N=16$.
(k)

$\beta=2, N=16$.
(1)

$\beta=3, N=16$.


C


D


E


F


G


H


I


J


L


PROB. Sp23-F.7. Shown below on the left are twelve pole-zero plots for $H(z)$, labeled A through L. Shown below on the right are the corresponding magnitude responses, $\left|H\left(e^{j \hat{\omega}}\right)\right|$ plotted $v s \hat{\omega}$, but in a scrambled order. Match each magnitude response to its corresponding pole-zero plot. Indicate answers by writing a letter (from A through L) into each answer box.


GEORGIA INSTITUTE OF TECHNOLOGY
SCHOOL of ELECTRICAL AND COMPUTER ENGINEERING

ECE 2026 Spring 2023
Final Exam
April 28, 2023


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## PROB. Sp23-F.1.

Suppose that the following equation is true, for all time $t$ :

$$
t_{0} \cos \left(2 \pi f_{0}\left(t-t_{0}\right)\right)=\sqrt{2} \cos \left(2 \pi f_{0} t\right)+\sqrt{2} \sin \left(2 \pi f_{0} t\right)
$$

(a) As a function of the unspecified parameters $f_{0}$ and $t_{0}$, the corresponding phasor equation is:

(b) Find the smallest positive values for $f_{0}$ and $t_{0}$ so that the given equation is true, for all time $t$ :


In polar form, the phasor equation simplifies to:

$$
t_{0} e^{-j 2 \pi f_{0} t_{0}}=2 e^{-j 0.25 \pi}
$$

- Equate magnitudes $\Rightarrow\left|t_{0} e^{-j 2 \pi f_{0} t_{0}}\right|=\left|2 e^{-j 0.25 \pi}\right| \Rightarrow t_{0}=2$
- Equate phases $\Rightarrow 2 \pi f_{0} t_{0}=0.25 \pi \Rightarrow f_{0}=\frac{0.25 \pi}{2 \pi t_{0}}=\frac{1}{8 t_{0}}=\frac{1}{16}$

PROB. Sp23-F.2. $\quad$ Let $x[n]=-2 \delta[n]+2 \delta[n-2]+\delta[n-4]$ be the input to an LTI filter whose frequency response is $H\left(e^{j \hat{\omega}}\right)=e^{-j \hat{\omega}}(3+4 \cos (\hat{\omega}))$, resulting in an output $y[n]$.

(a) ${ }^{\top} \square^{\mathrm{F}}$ (True or False): The system is FIR.
(b) The impulse response $h[n]$ satisfies:

$h[2]=2$.
(c) The filter output $y[n]$ satisfies:

$$
y[0]=-4
$$

$$
y[1]=-6
$$

$$
\begin{array}{rrrrrrr}
-4 & 0 & 4 & 0 & 2 & & \\
& -6 & 0 & 6 & 0 & 3 & \\
+ & & -4 & 0 & 4 & 0 & 2 \\
\hline
\end{array}
$$

$$
y[2]=0
$$

$$
y[3]=6
$$

$$
y[4]=6
$$

$$
y[5]=3
$$

$$
y[6]=2
$$

$$
y[7]=0
$$

## PROB. Sp23-F.3.

Suppose that a continuous-time signal $x(t)=8 \cos (2000 \pi t) \sin (4000 \pi t)$ is fed to an ideal sampling/ filtering/reconstruction system, producing an output $y(t)$, as shown below:


$$
f_{\max }=3000 \mathrm{~Hz}
$$

The difference equation for the FIR filter is $y[n]=x[n]+B x[n-1]+x[n-2]$.
The parameter $B$ and the sampling rate $f_{s}$ are unspecified and different in each part below.
(a) How big must $f_{s}$ be to avoid aliasing?

(b) If a sampling rate of $f_{s}=12000 \mathrm{~Hz}$ results in a D-to-C output $y(t)$ that is periodic with fundamental frequency $f_{0}=3000 \mathrm{~Hz}$, then it must be that:

$$
B=\boxed{-\sqrt{3}}
$$

To have $f_{0}=3 \mathrm{kHz}$, the 1 kHz input must be nulled
With $f_{s}=12 \mathrm{kHz}$, the frequency to null is $\hat{\omega}=\pi / 6$

$$
\Rightarrow B=-2 \cos (\pi / 6)=-\sqrt{3}
$$

(c) When $f_{s}=6000 \mathrm{~Hz}$ and $B=2$, the D-to-C output has the standard form $y(t)=A \cos \left(2 \pi f_{0} t+\varphi\right)$, where:

With $f_{s}=6 \mathrm{kHz}$, the 3 kHz input $4 \sin (6000 \pi t)$ aliases to zero!


$$
\Rightarrow x[n]=4 \sin \left(\frac{\pi}{3} n\right)
$$

$$
f_{0}=\underbrace{}_{>0} \mathrm{~Hz}
$$

$$
\begin{aligned}
& \text { But } \begin{aligned}
H\left(e^{j \pi / 3}\right) & =1+2 e^{-j \pi / 3}+e^{-j 2 \pi / 3} \\
& =3 e^{-j \pi / 3} \\
\Rightarrow y[n]= & 12 \sin \left(\frac{\pi}{3} n-\frac{\pi}{3}\right) \\
\Rightarrow y(t)= & y\left[f_{s} t\right]
\end{aligned} \quad=12 \sin \left(2000 \pi t-\frac{\pi}{3}\right) \\
& \\
& =12 \cos \left(2000 \pi t-\frac{5 \pi}{6}\right)
\end{aligned}
$$

PROB. Sp23-F.4. Shown below are ten different impulse responses $h[n]$, labeled A through J, where $h[n]$ is plotted for times $n \in\{-5, \ldots, 40\}$ only:

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$$
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\end{equation*}
$$

(ii)


$$
H_{2}\left(e^{j \hat{\omega}}\right)=\frac{e^{-j 10 \hat{\omega}}}{1+0.9 e^{-j \hat{\omega}}}
$$

(iii)


$$
H_{3}\left(e^{j \hat{\omega}}\right)=\frac{1+0.9 e^{-j 10 \hat{\omega}}}{1-0.9 e^{-j \hat{\omega}}}
$$

(iv)


$$
H_{4}\left(e^{j \hat{\omega}}\right)=\frac{1-0.9 e^{-j 10 \hat{\omega}}}{1-0.9 e^{-j \hat{\omega}}}
$$



$$
\begin{equation*}
H_{5}\left(e^{j \hat{\omega}}\right)=2 e^{-j 7 \hat{\omega}} \frac{\sin (2.5 \hat{\omega})}{\sin (0.5 \hat{\omega})} \cos (5 \hat{\omega}) \tag{v}
\end{equation*}
$$



$$
\begin{equation*}
H_{6}\left(e^{j \hat{\omega}}\right)=2 j e^{-j 7 \hat{\omega}} \frac{\sin (2.5 \hat{\omega})}{\sin (0.5 \hat{\omega})} \sin (5 \hat{\omega}) \tag{vi}
\end{equation*}
$$



$$
\begin{equation*}
H_{7}\left(e^{j \hat{\omega}}\right)=\frac{0.5}{1-0.9 e^{-j(\hat{\omega}-0.1 \pi)}}+\frac{0.5}{1-0.9 e^{-j(\hat{\omega}+0.1 \pi)}} \tag{vii}
\end{equation*}
$$

(viii)


$$
H_{8}\left(e^{j \hat{\omega}}\right)=\frac{0.5}{1-0.9 e^{-j(\hat{\omega}-0.05 \pi)}}+\frac{0.5}{1-0.9 e^{-j(\hat{\omega}+0.05 \pi)}}
$$

(ix)


$$
H_{9}\left(e^{j \hat{\omega}}\right)=e^{-j 12 \hat{\omega}} \frac{\sin (2.5 \hat{\omega})}{\sin (0.5 \hat{\omega})}
$$

(x)

$H_{10}\left(e^{j \hat{\omega}}\right)=e^{-j 14.5 \hat{\omega}} \frac{\sin (5 \hat{\omega})}{\sin (0.5 \hat{\omega})}$.

## PROB. Sp23-F.5.

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- The third system has impulse response: $\quad h_{3}[n]=(0.8)^{n} u[n]$.
(a)

(True or False): The first system is FIR.
(b) $\stackrel{\top}{\mathrm{X}}^{\mathrm{T}}$ (True or False): The second system is FIR.
(c) $\square^{\top} \stackrel{\mathrm{F}}{\mathrm{X}}^{\mathrm{F}}$ (True or False): The third system is FIR.
(d) If the difference equation relating the overall output to the overall input is:

$$
y[n]=x[n]-x[n-1], \quad \Rightarrow H(z)=1-z^{-1}
$$

then it must be that:

Solve $H(z)=H_{1}(z) H_{2}(z) H_{3}(z)$ for $H_{2}(z)$ :

$$
b_{0}=\square
$$

$$
\begin{aligned}
\Rightarrow H_{2}(z) & =\frac{H(z)}{H_{1}(z) H_{3}(z)}=\frac{1-z^{-1}}{\left(\frac{1-0.8 z^{-1}}{1+0.8 z^{-1}}\right)\left(\frac{1}{1-\theta .8 z^{-1}}\right)} \\
& =\left(1+0.8 z^{-1}\right)\left(1-z^{-1}\right) \\
& =1-0.2 z^{-1}-0.8 z^{-2}
\end{aligned}
$$

PROB. Sp23-F.6. Let $x[n]=\delta[n]+\beta \delta[n-1]+\delta[n-2]$.
Shown below are plots of $\quad\left|\sum_{n=0}^{2} x[n] e^{-j k 2 \pi n / N}\right|$ versus $k \in\{0, \ldots N-1\}$, labeled A through L. (The axes are not labeled, only the shapes matter.)
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(e) $\square$ $\beta=0, N=15$.
(f)

$\beta=1, N=15$.
(g)

$\beta=2, N=15$.
(h)

$\beta=3, N=15$.
(i)

$\beta=0, N=16$.
(j)

$\beta=1, N=16$.
(k) E
$\beta=2, N=16$.
(1)

$\beta=3, N=16$.


PROB. Sp23-F.7. Shown below on the left are twelve pole-zero plots for $H(z)$, labeled A through L. Shown below on the right are the corresponding magnitude responses, $\left|H\left(e^{j \hat{\omega}}\right)\right|$ plotted $v s \hat{\omega}$, but in a scrambled order. Match each magnitude response to its corresponding pole-zero plot. Indicate answers by writing a letter (from A through L) into each answer box.


