ECE 3084

QUIZ 2

School of Electrical and Computer Engineering Georgia Institute of Technology April 2, 2015

Name: _____

- 1. The quiz is closed book, except for one 2-sided sheet of handwritten notes.
- 2. Turn off your phone and put it away. No tablets/laptops/WiFi/etc. Calculators are OK.
- 3. Final answers must be entered into the answer box.
- 4. Correct answers must be accompanied by concise justifications to receive full credit.
- 5. Do not attach additional sheets. If necessary, use the back of the previous page.

| Problem | Points | Score |
|---------|--------|-------|
| 1 | 20 | |
| 2 | 20 | |
| 3 | 20 | |
| 4 | 20 | |
| 5 | 20 | |
| TOTAL: | 100 | |

PROBLEM 1. (20 points) — Be sure to label both axes, i.e. label important frequencies and amplitudes!

The left-hand side of the figure below shows a continuous-time sinc signal $s(t) = \frac{\sin(2\pi t)}{(\pi t)}$ being fed to an AM *modulator* with carrier frequency 40 Hz, resulting in a modulated signal x(t):



[This AM modulator differs from the usual because it multiplies by $\sin(\cdot)$ instead of $\cos(\cdot)$.] The right-hand side shows a *demodulator* consisting of another multiplication by $\sin(\cdot)$, resulting in the signal w(t), followed by an ideal LPF with an unspecified gain and cutoff frequency.

(a) Sketch $X(j\omega)$, the Fourier transform of the modulated signal x(t).

(b) Sketch $W(j\omega)$, the Fourier transform of the signal w(t).

- (c) $\sum_{i.e.}^{\text{YES}}$ No Can the LPF be designed so that its output is the same as the modulator input, *i.e.* to achieve y(t) = s(t)?
- (d) If YES, sketch below the frequency response $H(j\omega)$ of the LPF that results in y(t) = s(t); If No, explain below why this is impossible:

PROBLEM 2. (20 points)

Suppose that the continuous-time sinc signal $x(t) = \frac{\sin(202\pi t)}{(202\pi t)}$ is sampled at an unspecified sampling rate f_s , and that the samples are immediately fed to an ideal D-to-C converter (with the same f_s parameter), producing the continuous-time output signal y(t), as shown below:



- (a) The zero-th sample is x[0] =
- (b) In order for the D-to-C converter to reconstruct the original signal (i.e., to achieve y(t) = x(t)), the sampling frequency must satisfy: $f_s >$ Hz.
- (c) In the space below, carefully sketch the output Fourier transform $Y(j\omega)$ when $f_s = 200$ Hz, carefully labeling important frequencies and amplitudes:



(d) There are an infinite number of sampling frequencies f_s such that the output signal will be $y(t) = \sin(\pi f_s t)/(\pi f_s t)$. Name any three:



PROBLEM 3. (20 points)

(a) The Laplace transform of $x(t) = (t^2 - 1)u(t - 2)$ is X(s) =

(b) The inverse Laplace transform of $X(s) = \frac{s+3}{s(s+0.5)}$ is $x(t) = \left| ()u(t) \right|$.

(c) Find the differential equation that relates the output y(t) and input x(t) of an LTI system whose transfer function is $H(s) = \frac{\pi + s}{s(s+1)}$.

(d) The impulse response h(t) of the system from part (c) approaches the constant $h(\infty) =$ in the limit as time grows large.

PROBLEM 4. (20 points)

Subject to the initial conditions $x(0^-) = \dot{x}(0^-) = 1$, the following differential equation:

$$\frac{d^2}{dt^2}x(t) + 9x(t) = 0,$$

has the following solution:

$$x(t) = (A + B\cos(Ct) + D\sin(Et) + Ft)u(t),$$

where:



PROBLEM 5. (20 points)

An LTI system (zero initial conditions) with input x(t) and output y(t) obeys the following differential equation:

$$3\frac{d^2}{dt^2}y(t) = 2x(t) - 120y(t) - 12\frac{d}{dt}y(t).$$

(a) Circle one: The system is [overdamped] [underdamped] [critically damped]?



(d) Its damping ratio is
$$\zeta =$$

(e) The *impulse* response of this system is $h(t) = Ae^{-Bt}\sin(Ct)$, where:

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| Table of Fourier Transform Pairs | | | | |
|----------------------------------|--|--|--|--|
| Signal Name | Time-Domain: $x(t)$ | Frequency-Domain: $X(j\omega)$ | | |
| Right-sided exponential | $e^{-at}u(t)$ $(a > 0)$ | $\frac{1}{a+j\omega}$ | | |
| Left-sided exponential | $e^{bt}u(-t) (b>0)$ | $\frac{1}{b-j\omega}$ | | |
| Square pulse | [u(t + T/2) - u(t - T/2)] | $\frac{\sin(\omega T/2)}{\omega/2}$ | | |
| "sinc" function | $\frac{\sin(\omega_0 t)}{\pi t}$ | $[u(\omega + \omega_0) - u(\omega - \omega_0)]$ | | |
| Impulse | $\delta(t)$ | 1 | | |
| Shifted impulse | $\delta(t-t_0)$ | $e^{-j\omega t_0}$ | | |
| Complex exponential | $e^{j\omega_0 t}$ | $2\pi\delta(\omega-\omega_0)$ | | |
| General cosine | $A\cos(\omega_0 t + \phi)$ | $\pi A e^{j\phi} \delta(\omega - \omega_0) + \pi A e^{-j\phi} \delta(\omega + \omega_0)$ | | |
| Cosine | $\cos(\omega_0 t)$ | $\pi\delta(\omega-\omega_0)+\pi\delta(\omega+\omega_0)$ | | |
| Sine | $\sin(\omega_0 t)$ | $-j\pi\delta(\omega-\omega_0)+j\pi\delta(\omega+\omega_0)$ | | |
| General periodic signal | $\sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$ | $\sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - k\omega_0)$ | | |
| Impulse train | $\sum_{n=-\infty}^{\infty} \delta(t - nT)$ | $\frac{2\pi}{T}\sum_{k=-\infty}^{\infty}\delta(\omega - 2\pi k/T)$ | | |

| Table of Fourier Transform Properties | | | | |
|---------------------------------------|-------------------------|---|--|--|
| Property Name | Time-Domain $x(t)$ | Frequency-Domain $X(j\omega)$ | | |
| Linearity | $ax_1(t) + bx_2(t)$ | $aX_1(j\omega) + bX_2(j\omega)$ | | |
| Conjugation | $x^*(t)$ | $X^*(-j\omega)$ | | |
| Time-Reversal | x(-t) | $X(-j\omega)$ | | |
| Scaling | f(at) | $rac{1}{ a }X(j(\omega/a))$ | | |
| Delay | $x(t-t_d)$ | $e^{-j\omega t_d}X(j\omega)$ | | |
| Modulation | $x(t)e^{j\omega_0 t}$ | $X(j(\omega-\omega_0))$ | | |
| Modulation | $x(t)\cos(\omega_0 t)$ | $\frac{1}{2}X(j(\omega-\omega_0)) + \frac{1}{2}X(j(\omega+\omega_0))$ | | |
| Differentiation | $\frac{d^k x(t)}{dt^k}$ | $(j\omega)^k X(j\omega)$ | | |
| Convolution | x(t) * h(t) | $X(j\omega)H(j\omega)$ | | |
| Multiplication | x(t)p(t) | $\frac{1}{2\pi}X(j\omega)*P(j\omega)$ | | |

Date 14-Sept-2013

| Table of Laplace Transform Pairs | | | | |
|----------------------------------|-------------------------------|---|--|--|
| Signal Name | Time-Domain: $x(t)$ | Laplace-Domain: $X(s)$ | | |
| Impulse | $\delta(t)$ | 1 | | |
| Delayed Impulse | $\delta(t-t_0), \ t_0 \ge 0$ | e^{-st_0} | | |
| Step | u(t) | $\frac{1}{s}$ | | |
| Rectangular Pulse | u(t) - u(t - T), T > 0 | $\frac{1 - e^{-sT}}{s}$ | | |
| Ramp | tu(t) | $\frac{1}{s^2}$ | | |
| Polynomial | $t^k u(t), \ k \ge 0$ | $\frac{k!}{s^{k+1}}$ | | |
| Exponential | $e^{-at}u(t)$ | $\frac{1}{s+a}$ | | |
| Polynomial \times Exponential | $t^k e^{-at} u(t)$ | $\frac{k!}{(s+a)^{k+1}}$ | | |
| Cosine | $\cos(\omega_0 t)u(t)$ | $\frac{s}{s^2 + \omega_0^2}$ | | |
| Sine | $\sin(\omega_0 t)u(t)$ | $\frac{\omega_0}{s^2 + \omega_0^2}$ | | |
| Exponential \times Cosine | $e^{-at}\cos(\omega_0 t)u(t)$ | $\frac{s+a}{(s+a)^2+\omega_0^2}$ | | |
| Exponential \times Sine | $e^{-at}\sin(\omega_0 t)u(t)$ | $\frac{\omega_0}{(s+a)^2 + \omega_0^2}$ | | |

| Table of Laplace Transform Properties | | | | |
|---------------------------------------|--|--|--|--|
| Property Name | Time-Domain: $x(t)$ | Laplace-Domain: $X(s)$ | | |
| Linearity | $\alpha x_1(t) + \beta x_2(t)$ | $\alpha X_1(s) + \beta X_2(s)$ | | |
| Right-Shift | $x(t-t_0), \ t_0 \ge 0$ | $e^{-st_0}X(s)$ | | |
| Time Scaling | x(at), a > 0 | $\frac{1}{a}X\left(\frac{s}{a}\right)$ | | |
| First Derivative | $\dot{x}(t) = rac{d}{dt}x(t)$ | sX(s) - x(0) | | |
| Second Derivative | $\ddot{x}(t) = \frac{d^2}{dt^2}x(t)$ | $s^2 X(s) - s x(0) - \dot{x}(0)$ | | |
| Integration | $\int_0^t x(\tau) d\tau$ | $\frac{X(s)}{s}$ | | |
| Modulation | $x(t)e^{at}$ | X(s-a) | | |
| Convolution | x(t) * h(t) | X(s)H(s) | | |
| Final Value Theorem | $\lim_{t \to \infty} \overline{x(t)}, \text{ (if limit exists)}$ | $\lim_{s \to 0} sX(s)$ | | |
| Initial Value Theorem | $\lim_{t \to 0} x(t), \text{ (if limit exists)}$ | $\lim_{s \to \infty} sX(s)$ | | |