

ECE 3084

QUIZ 2

SCHOOL OF ELECTRICAL AND COMPUTER ENGINEERING
GEORGIA INSTITUTE OF TECHNOLOGY
NOVEMBER 20, 2018

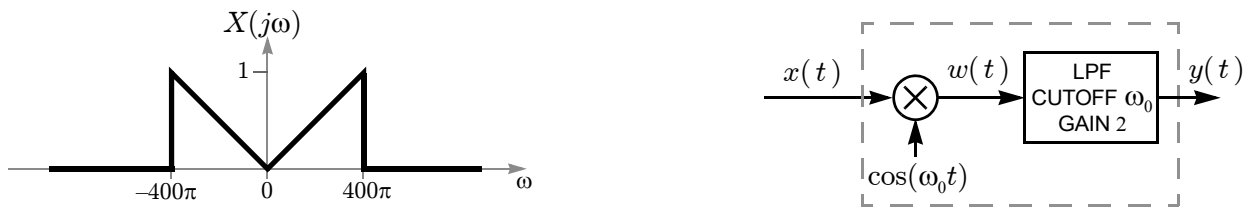
Name: _____

1. The quiz is closed book, closed notes, except for two 2-sided sheet of handwritten notes.
2. Turn off your phone and put it away. No tablets/laptops/WiFi/etc. No calculators.
3. Final answers must be entered into the answer box.
4. Correct answers *must be accompanied by concise justifications* to receive full credit.
5. Do not attach additional sheets. If necessary, use the back of the previous page.

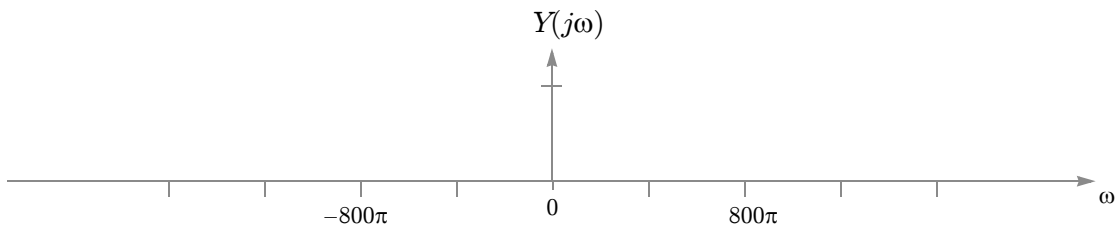
Problem	Points	Score
1	20	
2	20	
3	20	
4	20	
5	20	
TOTAL:	100	

PROBLEM 1.

Suppose a signal $x(t)$ with the FT $X(j\omega)$ shown below is passed as an input into the illustrated AM demodulator (dashed box), producing an output $y(t)$:



- (a) Sketch the Fourier transform $Y(j\omega)$ of the demodulator output for the special case when the demodulator carrier frequency is $\omega_0 = 800\pi$ (labeling amplitudes carefully!):



- (b) In order for the demodulator output $y(t)$ to be a sinc function, it must be that $\omega_0 = \boxed{} \text{ rad/s.}$

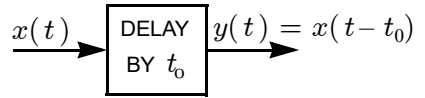
- (c) Using the value of ω_0 from part (b), the demodulator output has the form $y(t) = \frac{B \sin(Ct)}{t}$, where:

$$B = \boxed{}$$

$$C = \boxed{}$$

PROBLEM 2.

Suppose the signal $x(t) = 30\cos(15\pi t) - 84\sin(15\pi t)$ is fed as an input into a delay-by- t_0 system, producing the output $y(t) = x(t - t_0)$, as shown in the figure. Let $\omega_0 = 15\pi$.



- (a) The complex envelope of the input $x(t)$, with respect to ω_0 , is

$$\tilde{x}(t) = \boxed{\phantom{\frac{1}{s^2} \left(\frac{1}{s} + \frac{1}{s+1} \right)}}.$$

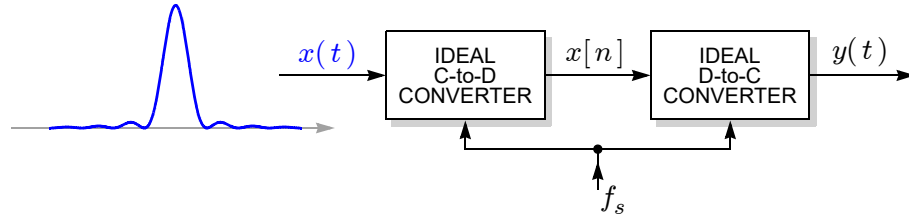
- (b) Find the *smallest** positive value for the delay $t_0 > 0$ so that the I & Q components of the delay output (with respect to ω_0) are $y_I(t) = 84$ and $y_Q(t) = -30$, respectively.

*there are multiple positive values of t_0 that do the job, but only one of them is the smallest

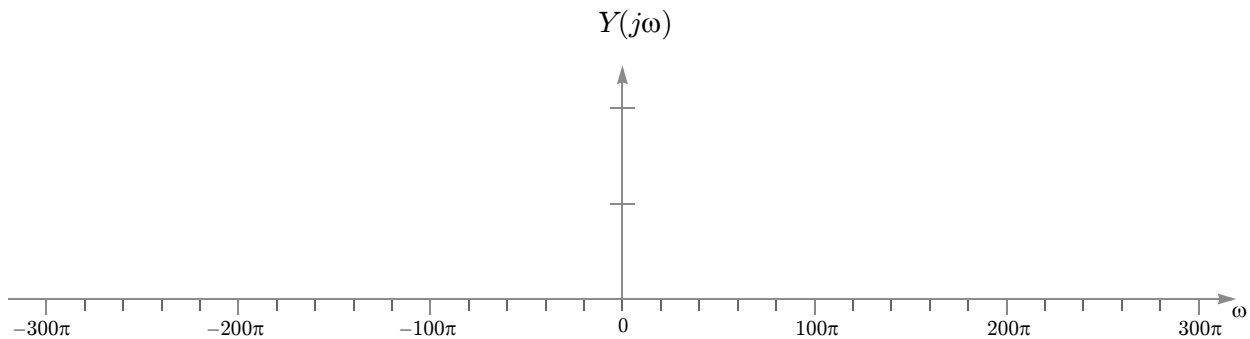
$$t_0 = \boxed{}$$

PROBLEM 3.

Suppose that the continuous-time sinc-squared signal $x(t) = \left(\frac{\sin(80\pi t)}{\pi t}\right)^2$ is sampled at an unspecified sampling rate f_s , and that the samples are immediately fed to an ideal D-to-C converter (or ideal DAC, with the same f_s parameter), producing the continuous-time output signal $y(t)$, as shown below:



- (a) The zero-th sample is $x[0] = \boxed{}$.
- (b) In order for the D-to-C converter to reconstruct the original signal (i.e., to achieve $y(t) = x(t)$), the sampling frequency must satisfy: $f_s > \boxed{}$ Hz.
- (c) In the space below, carefully sketch the output Fourier transform $Y(j\omega)$ when $f_s = 140$ Hz, *carefully labeling important amplitudes*:



PROBLEM 4.

Consider a signal $x(t)$ that obeys the following differential equation:

$$\frac{d^2}{dt^2} x(t) - 9x(t) = 0,$$

where one initial condition is specified, namely $x(0) = 2$,
while the other initial condition $\dot{x}(0)$ is not specified and may be nonzero.

If the solution to this differential equation has the form $x(t) = Be^{-Ct}u(t)$, where $C > 0$,
then it must be that:

$$\dot{x}(0) = \boxed{}$$

$$B = \boxed{}$$

$$C = \boxed{}$$

PROBLEM 5.

An LTI system (initially at rest, zero initial conditions) with input $x(t)$ and output $y(t)$ obeys the following differential equation:

$$\frac{d^2}{dt^2} y(t) = 2x(t) - 9y(t) - 4\frac{d}{dt} y(t).$$

(a) Circle one: The system is [undamped][overdamped][underdamped][critically damped]?

(b) Circle one: It most resembles a [LPF][HPF][BPF][notch] filter.

(c) Its d.c. gain is $H_0 =$.

(d) Its natural frequency is $\omega_n =$ (in rad/s).

(e) Its damping ratio is $\zeta =$.

(f) The *impulse* response of this system is $h(t) = Ae^{-Bt}\sin(Ct)$, where:

$A =$

$B =$

$C =$

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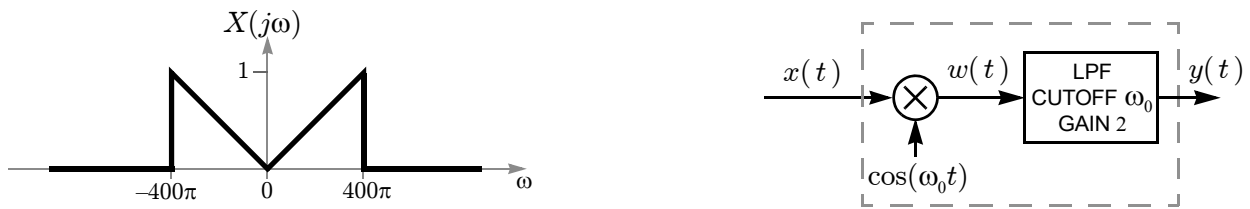
Name: ANSWER KEY

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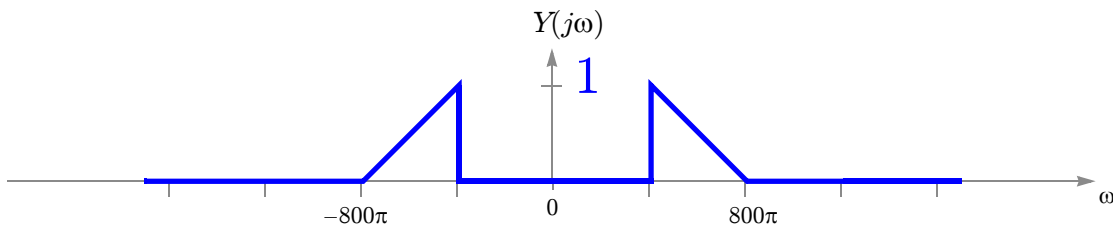
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PROBLEM 1.

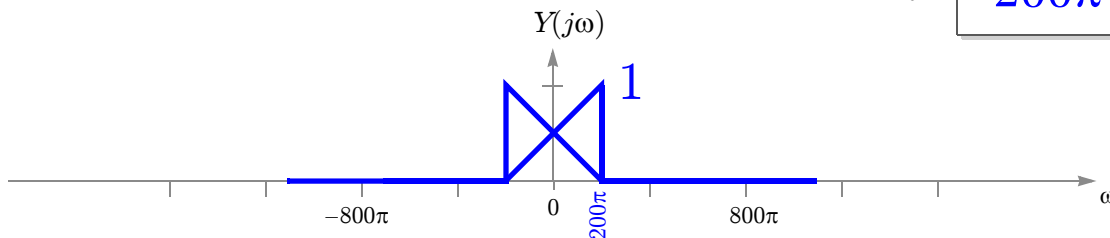
Suppose a signal $x(t)$ with the FT $X(j\omega)$ shown below is passed as an input into the illustrated AM demodulator (dashed box), producing an output $y(t)$:



- (a) Sketch the Fourier transform $Y(j\omega)$ of the demodulator output for the special case when the demodulator carrier frequency is $\omega_0 = 800\pi$ (labeling amplitudes carefully!):



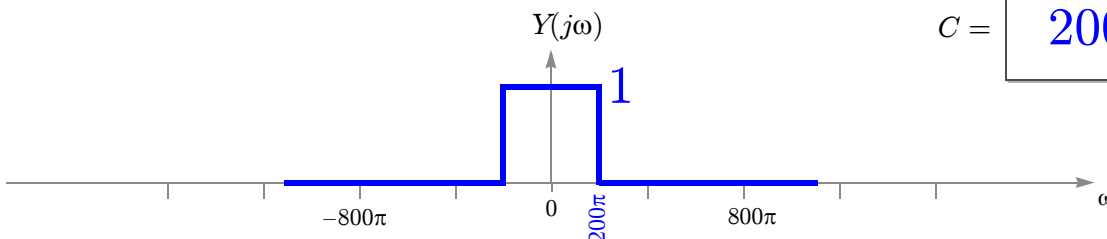
- (b) In order for the demodulator output $y(t)$ to be a sinc function, it must be that $\omega_0 =$ 200π rad/s.



- (c) Using the value of ω_0 from part (b), the demodulator output has the form $y(t) = \frac{B \sin(Ct)}{t}$, where:

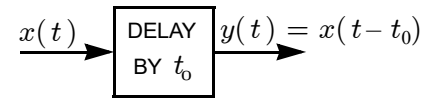
$$B = \frac{1}{\pi}$$

$$C = 200\pi$$



PROBLEM 2.

Suppose the signal $x(t) = 30\cos(15\pi t) - 84\sin(15\pi t)$ is fed as an input into a delay-by- t_0 system, producing the output $y(t) = x(t - t_0)$, as shown in the figure. Let $\omega_0 = 15\pi$.



- (a) The complex envelope of the input $x(t)$, with respect to ω_0 , is

$$\tilde{x}(t) = 30 + 84j.$$

- (b) Find *the smallest** positive value for the delay $t_0 > 0$ so that the I & Q components of the delay output (with respect to ω_0) are $y_I(t) = 84$ and $y_Q(t) = -30$, respectively.

*there are multiple positive values of t_0 that do the job, but only one of them is the smallest

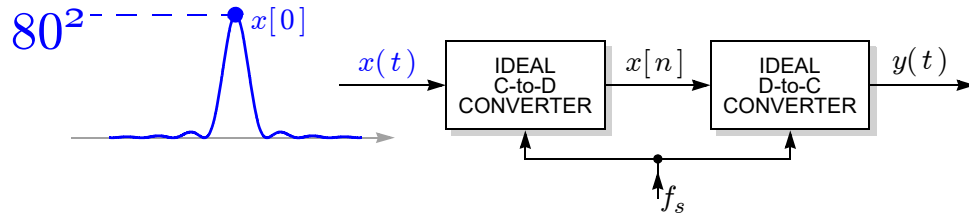
$$t_0 = \frac{1}{30}$$

$$\begin{aligned} y(t) &= x(t - t_0) \\ &= 30\cos(15\pi(t - t_0)) - 84\sin(15\pi(t - t_0)) \\ &= 30\cos(15\pi t - \theta) - 84\sin(15\pi t - \theta) \\ &= 84\cos(15\pi t) + 30\sin(15\pi t) \end{aligned}$$

$$\text{when } \theta = 15\pi t_0 = 0.5\pi \quad \Rightarrow \quad t_0 = \frac{0.5\pi}{15\pi} = \frac{1}{30}$$

PROBLEM 3.

Suppose that the continuous-time sinc-squared signal $x(t) = \left(\frac{\sin(80\pi t)}{\pi t}\right)^2$ is sampled at an unspecified sampling rate f_s , and that the samples are immediately fed to an ideal D-to-C converter (or ideal DAC, with the same f_s parameter), producing the continuous-time output signal $y(t)$, as shown below:



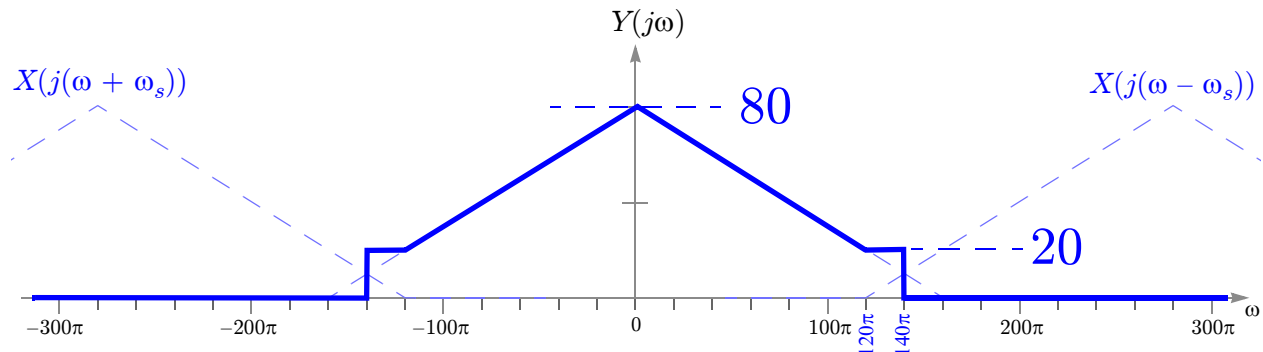
(a) The zero-th sample is $x[0] =$ 80^2 .

(b) In order for the D-to-C converter to reconstruct the original signal (i.e., to achieve $y(t) = x(t)$), the sampling frequency must satisfy:

$$2f_{\max} = 2(80) = 160$$

$f_s >$ 160 Hz.

(c) In the space below, carefully sketch the output Fourier transform $Y(j\omega)$ when $f_s = 140$ Hz, carefully labeling important amplitudes:



PROBLEM 4.

Consider a signal $x(t)$ that obeys the following differential equation:

$$\frac{d^2}{dt^2} x(t) - 9x(t) = 0,$$

where one initial condition is specified, namely $x(0) = 2$, while the other initial condition $\dot{x}(0)$ is not specified and may be nonzero.

If the solution to this differential equation has the form $x(t) = Be^{-Ct}u(t)$, where $C > 0$, then it must be that:

On the one hand, taking the LT of the given form yields:

$$X(s) = \frac{B}{s + C}$$

$$\begin{aligned} \dot{x}(0) &= \boxed{-6} \\ B &= \boxed{2} \\ C &= \boxed{3} \end{aligned}$$

On the other hand, taking the LT of both sides of the diff eqn yields:

$$s^2 X(s) - sx(0) - \dot{x}(0) - 9X(s) = 0$$

$$\Rightarrow X(s) = \frac{sx(0) + \dot{x}(0)}{s^2 - 9} = \frac{2(s + 0.5\dot{x}(0))}{(s + 3)(s - 3)}$$

Equating both forms, it must be that the numerator *cancels* the 2nd factor in the denominator:

$$\Rightarrow s + 0.5\dot{x}(0) = s - 3 \quad \Rightarrow \dot{x}(0) = -6$$

$$\Rightarrow X(s) \text{ reduces to } \frac{2}{s + 3} \quad \Rightarrow B = 2, C = 3.$$

PROBLEM 5.

An LTI system (initially at rest, zero initial conditions) with input $x(t)$ and output $y(t)$ obeys the following differential equation:

$$\frac{d^2}{dt^2} y(t) = 2x(t) - 9y(t) - 4\frac{d}{dt} y(t).$$

Taking LT, solving for $Y(s)/X(s) \Rightarrow H(s) = \frac{2}{s^2 + 4s + 9}$

Equate denom to $s^2 + 2\zeta\omega_n s + \omega_n^2$

$$\Rightarrow \omega_n = 3, \zeta = 2/3$$

- (a) Circle one: The system is [undamped] [overdamped] [underdamped] [critically damped]?

$$0 < \zeta < 1$$

- (b) Circle one: It most resembles a [LPF] [HPF] [BPF] [notch] filter.
numerator is constant

- (c) Its d.c. gain is $H_0 =$ $\frac{2}{9}$.
set $s = 0$

- (d) Its natural frequency is $\omega_n =$ 3 (in rad/s).

- (e) Its damping ratio is $\zeta =$ $\frac{2}{3}$.

- (f) The *impulse* response of this system is $h(t) = Ae^{-Bt}\sin(Ct)$, where:

$$A = \frac{H_0 \omega_n}{\sqrt{1 - \zeta^2}} = \frac{(2/9)(3)}{\sqrt{5}/3}$$

$$B = \zeta\omega_n = (2/3)(3)$$

$$C = \omega_d = \sqrt{1 - \zeta^2} \omega_n = \frac{\sqrt{5}}{3} (3)$$

$$A = \frac{2}{\sqrt{5}}$$

$$B = 2$$

$$C = \sqrt{5}$$